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Modeling Transceiver BER-OSNR Characteristic for QoT Estimation in Short-Reach Systems

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Abstract—A transceiver BER-OSNR model is validated and applied the Q-factor estimation for short-reach systems. Experiments using pluggable transceivers with commercial DSPs show that the modeling and estimation errors are less than 0.05 dB and 0.15 dB, respectively.

Index Terms—Transceiver model, QoT estimation, short-reach system

I. INTRODUCTION

The regional data center (DC) architecture replaces a few full-size DCs with several middle-size DCs to avoid space and power limitations and reduce the impact of natural disasters [1]. These distributed DCs are interconnected by high-speed and short-reach optical links where the transmission distance is less than a couple of hundred kilometers. The digital coherent system is a good candidate for these interconnections. Several open standards capable of 400G have been finalized [2], [3], and compatible modules are commercially available. To deploy an optical link, DC operators must estimate the quality of transmission (QoT), i.e., the signal bit error ratio (BER), and confirm that it is sufficient to run their services.

The existing QoT estimation method [4] first assesses the generalized signal-to-noise ratio (GSNR), which includes amplified spontaneous emission (ASE) noise and nonlinear interference (NLI) impairment caused by the fiber propagation. Then, it converts the obtained GSNR into the signal BER by consulting the transceiver characteristics that are described by the relationship between BER and optical signal-to-noise ratio (OSNR) in the back-to-back (B2B) setup. In the last decade, we have seen many studies on modeling the fiber propagation effects [5], including nonlinear effects, along with deploying the digital coherent system. However, the proper modeling of the transceiver characteristics remains basically untouched since phenomena occurring outside the transceiver are the dominant factors determining QoT in the traditional long-haul system. However, in short-reach systems like DC interconnections, the transceiver is one of the dominant impairments [6], [7], and thus, modeling the transceiver is essential for QoT estimation. Existing studies [6], [7] treat the transceiver impairment as additive Gaussian noise, but the BER-OSNR relation is still unclear. A decade-old study [8] briefly implied that the BER-OSNR relation can be modeled using additive Gaussian noise with a few other parameters. However, that study mainly focused on nonlinear effects, and

it is still unclear as to whether the implied model can be applied to current commercial systems. It is also unclear whether the implied model can be combined with other linear and nonlinear noise estimation methods to estimate the QoT accurately.

This paper revisits the transceiver model [8] and shows that this model can be applied to modern digital coherent systems equipped with commercial digital signal processors (DSPs). We evaluate the modeling error by using multiple transceivers with different modulation formats and baud rates, and the resulting root-mean-square error (RMSE) in Q-factor is less than 0.05 dB. We also demonstrate that the transceiver model combined with an open NLI noise estimation tool [4] can accurately estimate the QoT of short-reach systems. We set up multi-channel, multi-span short-reach systems comprising open and disaggregated transponders [9] and pluggable transceivers equipped with commercial DSPs. We compare the estimated and measured QoT for different span input powers and find that the estimation errors are less than 0.15 dB around the optimal power region.

II. MODELING OF THE TRANSCEIVER BER-OSNR CHARACTERISTIC

We use the following BER-OSNR model based on the previous one [8]:

$$\text{BER} = \Psi(\text{GSNR}), \quad (1)$$

$$\text{GSNR}^{-1} = \text{SNR}_{\text{ASE}}^{-1} + \text{SNR}_{\text{NLI}}^{-1} + \text{SNR}_{\text{TRX}}^{-1}, \quad (2)$$

$$\text{SNR}_{\text{ASE}} = \text{OSNR} \frac{\Delta f}{R_s \eta}, \quad (3)$$

where SNR_{ASE} is the signal-to-noise ratio (SNR) of ASE noise, SNR_{NLI} is the SNR of NLI noise, SNR_{TRX} is the SNR of the transceiver noise corresponding to the transceiver impairments, Δf is the OSNR measuring bandwidth, which is usually 12.5 GHz, R_s is the signal baud rate, η represents the deviations of the receiver filter from the ideal matched filter. If the receiver uses the ideal matched filter, η equals to 1. Otherwise, η is greater than 1. Function Ψ is mathematically determined from the modulation format and symbol mapping [10]; e.g., $\Psi(x) = \frac{1}{2} \text{erfc}(\sqrt{x/2})$ for DP-QPSK, $\Psi(x) = \frac{2}{3} \text{erfc}(\sqrt{\frac{3}{14}x})$ for DP-8QAM, $\Psi(x) =$

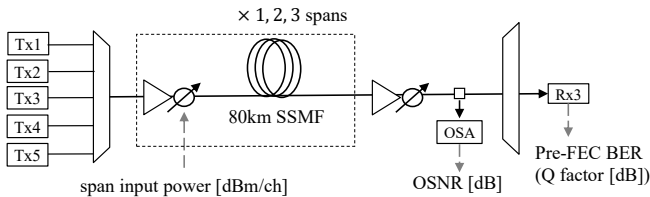


Fig. 3: Short-reach QoT estimation.

TABLE I: Q-factor changes caused by SNR 1.0 dB error

Total no. of spans	SNR_{TRX}	SNR_{ASE}	SNR_{NLI}
1	0.92	0.03	0.03
2	0.88	0.05	0.05
3	0.84	0.08	0.08

We keep the input power to the receiver at -7.0 dBm. We change the span input powers and measure the BER and OSNR. We estimate Q-factor using Eqs. (1) to (3) that requires OSNR, SNR_{NLI} , SNR_{TRX} , and η . As for SNR_{TRX} and η , we calculate them from preliminary B2B measurements (Fig. 1). As for OSNR, we extract amplifier noise figures from the input power and OSNR relation measured in the short-reach systems (Fig. 3). Then, we compute OSNR as $(\sum_i N_i h f_0 \Delta f / P_{\text{in},i})^{-1}$ where $P_{\text{in},i}$ is the input power to the i -th amplifier, N_i is the i -th amplifier's noise figure, h is Planck's constant, and f_0 is the signal frequency. As for SNR_{NLI} , we use two NLI estimation methods: GNPpy [4] and the Split-Step Fourier Method (SSFM). In the NLI calculation, we set the fiber nonlinearity coefficient to the typical value of 1.3 1/W/km.

Fig. 4 shows estimated and measured Q-factors in 1-span, 2-span, and 3-span setups. The optimal input power that maximizes the measured Q-factor is around 2.0 dBm/ch in all three cases. We can see the gap between estimation and measurement in the high power region, say more than 5.0 dBm/ch. We suppose that this deviations comes from using typical values of the fiber nonlinearity coefficient instead of an actual measured value. If we can use the coefficient that matches the measurements, the model calculation result agrees well with the measurements (the dashed lines in Fig. 4). Thus, the estimation error in the high-power region is due to the NLI estimation error and not the transceiver model. The estimation errors in the range of the optimal power ± 3.0 dBm/ch, i.e., from -1.0 dBm/ch to 5.0 dBm/ch, are as follows. RMSE with GNPpy of 1-span, 2-span, 3-span setup are 0.07 dB, 0.06 dB, 0.09 dB, respectively. Those with SSFM are 0.06 dB, 0.13 dB, 0.12 dB, respectively. Hence, regardless of the NLI estimator, the transceiver model well estimates the QoT of short-reach systems around the optimal power region.

Our quantitative Q-factor evaluations show that the accurate transceiver characteristic, i.e., the accurate value of SNR_{TRX} , is vital for QoT estimation. To assess the impact of estimation accuracy on the Q-factor, we compute the Q-factor change caused by a 1.0 dB error in SNR_{TRX} , SNR_{ASE} , or SNR_{NLI} , assuming other SNRs are accurate, at input power of 2.0 dBm/ch using Eqs. (1) to (3). Tab. I shows the results. The error in SNR_{TRX} has more impact than SNR_{ASE} and

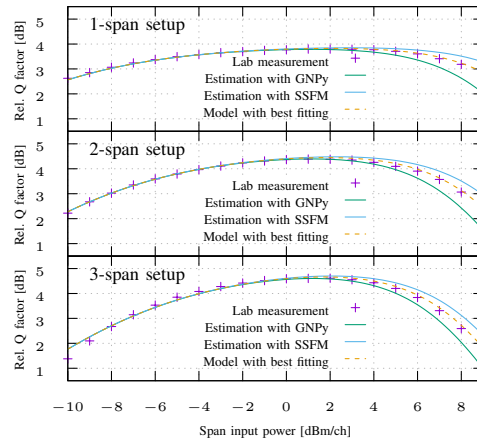


Fig. 4: Comparisons of Q-factor measurement and estimation.

SNR_{NLI} . If the number of spans is small, the impact of error in SNR_{ASE} and SNR_{NLI} is minor and could be negligible. Thus, in short-reach systems, accurately obtaining SNR_{TRX} is more critical than SNR_{ASE} and SNR_{NLI} .

V. CONCLUSION

We revisited a transceiver BER-OSNR characteristic model and validated it anew, showing that the modeling errors in Q-factor were less than 0.05 dB with modern commercial DSPs, and multiple modulation formats, and baud rates. Experiments showed that the transceiver model combined with GNPpy could reasonably estimate the QoT of real short-reach systems, including open and disaggregated transponders and pluggable transceivers equipped with commercial DSPs. The estimation errors were less than 0.15 dB around the optimal power region. Together with these results, our analysis of Q-factor changes caused by SNR changes highlights the importance of accurately modeling and acquiring the transceiver characteristic in short-reach systems.

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