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Handling anisotropic conductivities in the EEG forward problem with a symmetric formulation / Pillain, A., Rahmouni, L., Andriulli, F.. - In: PHYSICS IN MEDICINE & BIOLOGY. - ISSN 1361-6560. - 64:3(2019), p. 035022. [10.1088/1361-6560/aafaaf]

Availability:

This version is available at: 11583/2726910 since: 2019-03-03T20:35:51Z

Publisher:

NLM (Medline)

Published

DOI:10.1088/1361-6560/aafaaf

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Handling Anisotropic Conductivities in the EEG Forward Problem with a Symmetric Formulation

Axelle Pillain^a, Lyes Rahmouni^a, Francesco Andriulli^{b,a,*}

^a*Computational Electromagnetics Research Laboratory, IMT Atlantique, Brest, France*

^b*Politecnico di Torino, Turin, Italy*

Abstract

The electroencephalography (EEG) forward problem, the computation of the electric potential generated by a known electric current source configuration in the brain, is a key step of EEG source analysis. In this problem, it is often desired to model the anisotropic conductivity profiles of the skull and of the white matter. These profiles, however, cannot be handled by standard surface integral formulations and the use of volume finite elements is required. This paper addresses this issue. Leveraging on the representation theorem using an anisotropic fundamental solution, it proposes a modified symmetric formulation for solving the EEG forward problem by a surface integral equation which can take into account anisotropic conductivity profiles. A set of numerical results is presented to corroborate theoretical treatments and to show the impact of the proposed approach on both canonical and real case scenarios.

Keywords: Electroencephalography, EEG forward problem, anisotropy, Integral equations, Boundary Element Method

1. Introduction

Epileptic source localization from high resolution electroencephalographies (EEGs) is a fundamental step in the pre-surgical evaluation of focal epileptic patients that are refractory to pharmacological treatments and for whom surgical resection of the epileptic focus is considered [1–4]. In this brain imaging technique, starting from the electrical potential measured on the scalp, the brain current sources responsible for focal epilepsy are localized by solving an inverse source problem [5–7]. Solving this inverse problem requires the multiple solution of an EEG forward problem that provides, from known brain electrical current sources, the surface potential measured at the electrodes' locations [5, 8]. The solution of the forward problem must be computed with the highest possible precision [9–11] to increase the accuracy of the source localisation process.

In solving the EEG forward problem, spherical head models have been historically used since analytic solutions are available for them [12, 13]. However, modern techniques rely on the use of realistic head models that require a numerical solution, but

*Corresponding author

Email address: francesco.andriulli@imt-atlantique.fr (Francesco Andriulli)

Preprint submitted to Physics in Medicine and Biology

March 3, 2019

for which the accuracy of the forward EEG solution is largely improved [14–18]. Several methods can be used for numerically solving the EEG forward problem, including Finite Difference (FDM) [19, 20], Finite Element (FEM) [21] and Boundary Element methods (BEM) [22–25]. Boundary Element methods have been quite popular given that they require only surfacic discretization of the brain layers when compared to the other two techniques that rely on volume discretizations. These methods however are not panacea given that, in their standard incarnations, they cannot handle anisotropic conductivity profiles. Indeed, correct modeling of anisotropic conductivity profiles is quite important given the influence of white matter and skull anisotropic conductivities on source localization [26–32]. The reader should notice that anisotropic conductivity profiles can be naturally treated with FEM approaches. Even if the computational load can be reduced by resorting to transfer matrix, as in [33] or [34], a purely surfacic BEM method (with anisotropic modeling capabilities) could be desirable. Indeed, in solving the EEG forward problem, the integral methods which are based on an operator having an asymptotically smooth kernel can easily be used with a fast multipole method (FMM) [35] and this reduce their complexity to a quasi-linear one ($O(N \log(N))$, where N is the number of unknowns). Moreover, exploiting the reciprocity theorem as in [19, 36], further reduces the computational burden when solving the EEG inverse problem. Some steps in building an anisotropic BEM formulation have been presented in [37] where virtual sources are placed in the fibers to account for their anisotropic properties. Moreover, [38] proposed an interesting coordinate transform to handle constant anisotropy in a single conducting body. No details, however, were provided for the multi-compartments problem and the associated numerical solutions. Preliminary and partial results on an indirect anisotropic formulation for the EEG forward problem had been presented in the conference contributions [39, 40].

This work proposes a new approach to the problem that falls under the family of symmetric formulations as originally introduced in [24] and results in a surfacic discretized BEM formulation capable of handling constant piecewise homogeneous conductivity profiles including anisotropies for both nested and non-nested compartments. This is obtained by leveraging on the representation theorem for the Poisson equation, using an anisotropic fundamental solution. Instead of building a harmonic solution as is commonly done in building EEG integral formulations, our approach directly computes the unknown potential. This allows to enforce boundary conditions also across different anisotropic media. This approach also naturally permits to take into account non-nested domains as in [41], and it therefore renders possible further subdivisions of the compartments according to the variation of their conductivity tensors.

The article is organized as follows: Section 2 describes background material and sets the notations. Section 3 presents the new anisotropic symmetric integral equation. Section 4 focuses on the discretization of the BEM solution and its implementation. Section 5 complements all the theoretical developments with numerical results that show the effectiveness of the newly proposed method in both canonical and real case scenarios.

2. Background and Notation

Consider a domain Ω divided into N non overlapping subdomains Ω_i with Lipschitz boundary $\partial\Omega_i$ such that $\Omega = \bigcup_j^N \bar{\Omega}_i$. The external domain, $\mathbb{R}^3 \setminus \Omega$, is denoted with Ω_{N+1} .

We further define $\Gamma_{ij} = \bar{\Omega}_i \cap \bar{\Omega}_j$ the interface between Ω_i and Ω_j as well as ω_i the set of neighboring domains of Ω_i so that $\omega_i = \{j \leq N+1 | \bar{\Omega}_i \cap \bar{\Omega}_j \neq \emptyset\}$. Then ∂_{Ω_i} can be defined as $\partial_{\Omega_i} = \bigcup_{k \in I_i} \Gamma_k$ with $I_i = \{k | \Gamma_k = \Gamma_{ij}, j \in \omega_i\}$, the set of interfaces composing ∂_{Ω_i} . The normal \vec{n}_{ij} to the interface Γ_{ij} is oriented from Ω_i to Ω_j with $i < j$. Figure 1 shows a general decomposition of Ω into subdomains. Standard EEG BEM formulations often make use of nested domains [24], so that when a three layers geometry is chosen, the compartments represents the brain, the skull and the scalp. In this particular case, $\partial_{\Omega_i} = \Gamma_{i,i-1} \cup \Gamma_{i,i+1}$. We define the traces of a function g on a boundary Γ_i and of its conormal derivative as [42]

$$\gamma_{0i}^{\pm} g = g|_{\Gamma_i^{\pm}} \quad (1a)$$

$$\gamma_{1i}^{\pm} g = \vec{n} \cdot \bar{\sigma} \nabla g|_{\Gamma_i^{\pm}}. \quad (1b)$$

where $\bar{\sigma}$ is the conductivity tensor. Moreover $[\cdot]_k$ will denote the jump of a function across the surface Γ_k : $[g]_k = \gamma_{0k}^- g_k - \gamma_{0k}^+ g_k$ and $[\vec{n} \cdot \bar{\sigma} \nabla g]_k = \gamma_{1k}^- g_k - \gamma_{1k}^+ g_k$.

The EEG forward problem amounts at computing the electric potential V knowing the brain electric sources $f = \nabla \cdot j$ when the current j propagates in a medium of conductivity $\bar{\sigma}$ which is a real symmetric and positive definite matrix. Under standard quasi-static assumptions this calls for the solution of the Poisson's equation [43]

$$\nabla \cdot \bar{\sigma} \nabla V = f \quad (2)$$

with boundary conditions at each interface Γ_k

$$[V]_k = 0 \quad (3a)$$

$$[\vec{n} \cdot \bar{\sigma} \nabla V]_k = 0. \quad (3b)$$

The conditions above enforce the continuity of the potential and of its derivative between different compartments of the head. The conductivity $\bar{\sigma}$ is assumed to be piecewise homogeneous and potentially anisotropic, $\bar{\sigma}_i$ will denote the conductivity of the domain Ω_i . The source term $f = \nabla \cdot j$ is usually a linear combination of dipole sources f_j such that $f_j = q_j \cdot \nabla \delta_{r_{0j}}$ with q_j the dipole moment and r_{0j} the dipole position [44], [45].

3. A Surface Integral Formulation for Anisotropic Conductivity Profiles

The fundamental solution (Green's function) of (2) in an unbounded medium reads [46]

$$G_j(r, r') = \frac{1}{4\pi} \frac{1}{\sqrt{\det(\bar{\sigma}_j)} \sqrt{\bar{\sigma}_j^{-1}(r - r') \cdot (r - r')}} \quad (4)$$

where $\bar{\sigma}_j$ denotes the homogeneous conductivity tensor of the domain Ω_j . Decomposing the potential V into different contributions in each domains Ω_i , $V = \sum_{i=1}^N V_i$ such that $V_i(r) = V(r)$ if $r \in \Omega_i$ and 0 elsewhere, the representation theorem in Ω_i gives [42]

$$V_i(r) = \int_{\Omega_i} G_i(r, r') f(r') dr' + \int_{\partial\Omega_i} \bar{\sigma}_j \vec{n}' \cdot \nabla G_i(r, r') V(r') dr' - \int_{\partial\Omega_i} \bar{\sigma}_j \vec{n}' \cdot \nabla V(r') G_i(r, r') dr'. \quad (5)$$

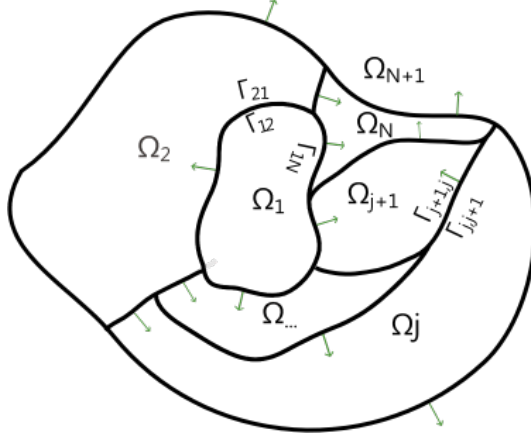


Figure 1: Decomposition of the domain Ω into subdomains Ω_i with boundary Γ_{ij} and normal \vec{n} .

Define the following integral operators

$$\mathcal{S}_i \nu(r) = \int_{\partial\Omega_i} G_i(r, r') \nu(r') dr' \quad \mathcal{S}_i : H^{-1/2}(\partial\Omega_i) \rightarrow H^1(\Omega) \quad (6)$$

and

$$\mathcal{D}_i \mu(r) = \int_{\partial\Omega_i} \mu(r') \vec{n}' \cdot (\vec{\sigma}_i \nabla G_i(r, r')) dr' \quad \mathcal{D}_i : H^{1/2}(\partial\Omega_i) \rightarrow H^1(\Omega). \quad (7)$$

Here and in the following H^s will denote the Sobolev space of order s . Then, since $\vec{\sigma}_i$ is symmetric, the representation theorem (5) reads

$$V_i(r) = \int_{\Omega_i} G_i(r, r') f(r') dr' + \mathcal{D}_i V(r) - \mathcal{S}_i \vec{n} \cdot \vec{\sigma}_i \nabla V(r'). \quad (8)$$

Note that, with our notations and normal orientation, [42]

$$\begin{cases} \gamma_{0i}^\pm \mathcal{S}_j = \sum_{k \in I_j} S_{ik, \Omega_j} & (9a) \\ \gamma_{1i}^\pm \mathcal{S}_j = \mp \frac{1}{2} + \sum_{k \in I_j} D_{ij, \Omega_j}^* & (9b) \end{cases} \quad \text{and} \quad \begin{cases} \gamma_{0i}^\pm \mathcal{D}_j = \mp \frac{1}{2} + \sum_{k \in I_j} D_{ij, \Omega_j} & (10a) \\ \gamma_{1i}^\pm \mathcal{D}_j = \sum_{k \in I_j} N_{ij, \Omega_j} & (10b) \end{cases}$$

with,

$$S_{ik,\Omega_j}\nu(r) = \int_{\Gamma_k} G_j(r, r')\nu(r')dr', \quad S_{ij} : H^{-1/2}(\Gamma_k) \rightarrow H^{1/2}(\Gamma_i), \quad (11a)$$

$$D_{ik,\Omega_j}^*\nu(r) = \int_{\Gamma_k} \nu(r')\vec{n} \cdot (\bar{\sigma}_j \nabla G_j(r, r'))dr', \quad D_{ij}^* : H^{-1/2}(\Gamma_k) \rightarrow H^{-1/2}(\Gamma_i), \quad (11b)$$

$$D_{ik,\Omega_j}\mu(r) = \int_{\Gamma_k} \mu(r')\vec{n}' \cdot (\bar{\sigma}_j \nabla G_j(r, r'))dr', \quad D_{ij} : H^{1/2}(\Gamma_k) \rightarrow H^{1/2}(\Gamma_i), \quad \text{and} \quad (11c)$$

$$N_{ik,\Omega_j}\mu(r) = \vec{n} \cdot \bar{\sigma}_i \nabla \int_{\Gamma_k} \mu(r')\vec{n}' \cdot (\bar{\sigma}_j \nabla G_j(r, r'))dr', \quad N_{ij} : H^{1/2}(\Gamma_k) \rightarrow H^{-1/2}(\Gamma_i). \quad (11d)$$

Now, let us consider $\Gamma_k = \overline{\Omega_i} \cap \overline{\Omega_j} \neq \emptyset$ with $i < j$ to express the interior problems and combine the obtained equations to enforce the boundary conditions at the interface Γ_k (see figures 2(a), 2(b), and 2(c)). The first step is to express also the interior problem (8) in Ω_j

$$V_j(r) = \int_{\Omega_j} G_j(r, r')f(r')dr' + \mathcal{D}_j V(r) - \mathcal{S}_j \vec{n} \cdot \bar{\sigma}_j \nabla V(r). \quad (12)$$

Then, decomposing $V_{j|\partial\Omega_j}$ and $V_{i|\partial\Omega_i}$ into their contributions on the different interfaces Γ_l with $l \in I_j$ or I_i respectively, such that $V_{j|\Gamma_l} = \mathcal{V}_l$ for $l \in I_j$ or $V_{i|\Gamma_l} = \mathcal{V}_l$ for $l \in I_i$ respectively, and taking the trace of (12) and of (8) we obtain

$$\gamma_{0_k}^+ V_j = \gamma_{0_k}^+ v_{dip_j} + \sum_{l \in I_j} D_{kl,\Omega_j} \mathcal{V}_l + \frac{1}{2} \mathcal{V}_k + \sum_{l \in I_j} S_{kl,\Omega_j} \vec{n} \cdot \bar{\sigma}_j \nabla \mathcal{V}_l. \quad (13)$$

$$\gamma_{0_k}^- V_i = \gamma_{0_k}^- v_{dip_i} + \sum_{l \in I_i} D_{kl,\Omega_i} \mathcal{V}_l + \frac{1}{2} \mathcal{V}_k - \sum_{l \in I_i} S_{kl,\Omega_i} \vec{n} \cdot \bar{\sigma}_i \nabla \mathcal{V}_l. \quad (14)$$

where to simplify the notations we introduced $v_{dip_j}(r) = \int_{\Omega_j} G_j(r, r')f(r')dr'$ the potential due to the source in Ω_j . Finally, enforcing the boundary condition (3a) by subtracting (13) to (14) we obtain

$$\begin{aligned} \sum_{l \in I_i} D_{kl,\Omega_i} \mathcal{V}_l - \sum_{l \in \omega_i} D_{kl,\Omega_j} \mathcal{V}_l + \sum_{l \in I_i} S_{kl,\Omega_i} \vec{n} \cdot \bar{\sigma}_i \nabla \mathcal{V}_l \\ - \sum_{l \in I_j} S_{kl,\Omega_j} \vec{n} \cdot \bar{\sigma}_j \nabla \mathcal{V}_l = -(\gamma_{0_k}^- v_{dip_i} - \gamma_{0_k}^+ v_{dip_j}) \end{aligned} \quad (15)$$

The same reasoning using the trace of the conormal derivative gives the two following interior problems in Ω_j and Ω_i

$$\gamma_{1_k}^- V_i = \gamma_{1_k}^- v_{dip_i} + \sum_{l \in I_i} N_{kl,\Omega_i} \mathcal{V}_l + \frac{1}{2} \vec{n} \cdot \bar{\sigma}_i \mathcal{V}_k - \sum_{l \in I_i} D_{kl,\Omega_i}^* \vec{n} \cdot \bar{\sigma}_i \nabla \mathcal{V}_l \quad (16)$$

and

$$\gamma_{1_k}^+ V_j = \gamma_{1_k}^+ v_{dip_j} + \sum_{l \in I_j} N_{kl,\Omega_j} \mathcal{V}_l + \frac{1}{2} \vec{n} \cdot \bar{\sigma}_j \mathcal{V}_k - \sum_{l \in I_j} D_{kl,\Omega_j}^* \vec{n} \cdot \bar{\sigma}_j \nabla \mathcal{V}_l. \quad (17)$$

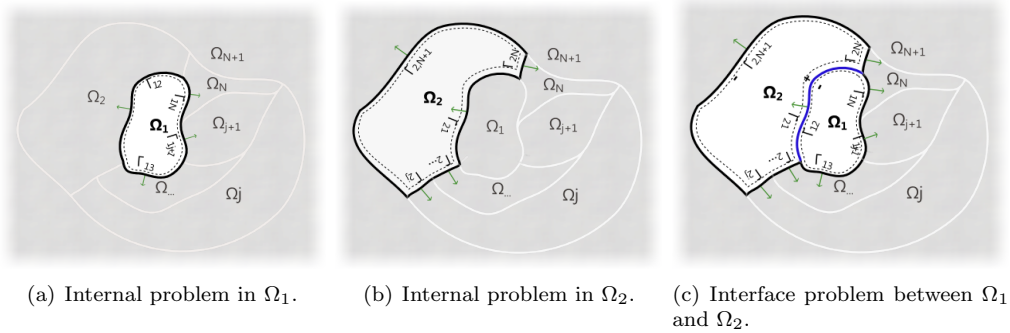


Figure 2: Conventions used in setting up the integral equations.

Given that at the interface Γ_k , $\frac{1}{2}\vec{n} \cdot \vec{\sigma}_i \mathcal{V}_k = \frac{1}{2}\vec{n} \cdot \vec{\sigma}_j \mathcal{V}_k$, the equations (16) and (17) can be combined using the second boundary condition (3b) to get a second integral equation

$$\begin{aligned} \sum_{l \in I_i} N_{kl, \Omega_i} \mathcal{V}_l - \sum_{l \in I_j} N_{kl, \Omega_j} \mathcal{V}_l + \sum_{l \in I_i} D_{kl, \Omega_i}^* \vec{n} \cdot \vec{\sigma}_i \nabla \mathcal{V}_l \\ - \sum_{l \in I_j} D_{kl, \Omega_j}^* \vec{n} \cdot \vec{\sigma}_j \nabla \mathcal{V}_l = -(\gamma_{1_k}^- v_{dip_i} - \gamma_{1_k}^+ v_{dip_j}). \end{aligned} \quad (18)$$

The equations (15) and (18) correspond to the symmetric formulation presented in [24]. However, the introduced formulation differs from it in its essence as the kernel of the involved integral operators is not the same. Moreover, the presenting reasoning generalizes the use of this formulation to fully inhomogeneous and anisotropic domains. As in the case of the standard symmetric formulation, at the external interface(s), only the boundary condition (3b) is necessary to be enforced, this means that for this(ese) interface(s), only (18) will be enforced. Indeed, (3b) at the outermost interface directly tells that the conormal derivative of the potential is zero here.

4. Discretization and Implementation of the Modified Symmetric Equation

To obtain a linear system to be solved, following the usual BEM strategy [42], the integral equations must be tested with a suitably chosen set of basis functions. Tessellating the geometry into N_t triangular cells t_i and N_v vertices v_i , this gives rise to the linear system

$$\mathbf{Z}\mathbf{a} = \mathbf{b} \quad (19)$$

where \mathbf{Z} contains the discrete version of the integral equation. Each entry of \mathbf{Z} is given by

$$[\mathbf{Z}]_{kl} = \int_{\mu_k^*} f_{t_k}(r') Z f_{e_l}(r') dr' \quad (20)$$

where Z is the operator of the integral equation under consideration, $\{f_{t_k}\}$ is the set of testing functions, and $\{f_{e_k}\}$ is the set of expansion functions whose supports are given by $\{\mu_{e_k}\}$. In our case, the unknowns \mathcal{V}_i and $\vec{n} \cdot \vec{\sigma}_j \nabla \mathcal{V}_i$ can be respectively discretized using,

as expansion functions, piecewise linear functions $P_{1k} \in H^{1/2}$ and piecewise constant functions $P_{0k} \in H^{-1/2}$. Such that,

$$\mathcal{V}_j \approx \sum_k \alpha_{j,k} P_{1k} \text{ and } \bar{\sigma}_i \nabla \mathcal{V}_j \approx \sum_k \beta_{j,k} P_{0k},$$

where $P_{0i}(r) = 1$ if $r \in t_i$ and 0 elsewhere and $P_{1i}(r) = 1$ if $r = v_i$ and linearly decreases to 0 on their support $\mu_{P_{1i}} = \{t_k | v_i \text{ is a vertex of } t_k\}$. The support of P_{0i} is given by $\mu_{P_{0i}} = t_i$.

Expansion and testing functions should comply with the operatorial mappings of the involved integral operators. The range of S_{ij} and D_{ij} is $H^{1/2}$ for which the dual space is $H^{-1/2}$. As $P_0 \in H^{-1/2}$ these functions can be used for testing equation (15). The operators involved in (18) are D_{ij}^* and N_{ij} whose range is $H^{-1/2}$. This means that pyramidal functions $P_1 \in H^{1/2}$ are suitable testing functions. This discretization corresponds to the one presented in [24]. Then the entry of the matrix \mathbf{Z} will be given by the discrete version of the operators $S_{ij,\Omega_k}, D_{ij,\Omega_k}, D_{ij,\Omega_k}^*$ and N_{ij,Ω_k} with

$$\begin{aligned} [\mathbf{S}_{ij,\Omega_k}]_{kl} &= \int_{\mu_{P_{0k}}} P_{0k} S_{ij,\Omega_k} P_{0l}(r) dr \\ [\mathbf{D}_{ij,\Omega_k}]_{kl} &= \int_{\mu_{P_{0k}}} P_{0k} D_{ij,\Omega_k} P_{1l}(r) dr \\ [\mathbf{D}_{ij,\Omega_k}^*]_{kl} &= \int_{\mu_{P_{1k}}} P_{1k} D_{ij,\Omega_k}^* P_{0l}(r) dr \\ [\mathbf{N}_{ij,\Omega_k}]_{kl} &= \int_{\mu_{P_{1k}}} P_{1k} N_{ij,\Omega_k} P_{1l}(r) dr. \end{aligned}$$

Note here that only surface basis functions are required. As such, only a surface mesh is necessary to implement the presented formulation. The separation of the considered geometry into different homogeneous anisotropic or isotropic compartments is however a necessary pre-processing step.

The proposed formulation can be interesting to model the skull anisotropy where in this layer, different conductivity tensors would correspond to different domains. Following this, as well as to fix the ideas and to simplify the implementation, we provide below the explicit form of the matrices in the case of the geometry depicted figure 3. In this simplified case, the system matrix reads

$$\mathbf{Z} = \begin{bmatrix} \mathbf{S} & \mathbf{D} \\ \mathbf{D}^* & \mathbf{N} \end{bmatrix},$$

with

$$\mathbf{A} = \begin{pmatrix} A_{11,\Omega_1} - A_{11,\Omega_2} & A_{12,\Omega_1} & -A_{13,\Omega_2} & -A_{14,\Omega_2} & 0 & 0 \\ A_{21,\Omega_1} & A_{22,\Omega_1} - A_{22,\Omega_3} & -A_{23,\Omega_3} & 0 & -A_{25,\Omega_3} & 0 \\ A_{31,\Omega_2} & -A_{32,\Omega_3} & A_{33,\Omega_2} - A_{33,\Omega_3} & A_{34,\Omega_2} & -A_{35,\Omega_3} & 0 \\ A_{41,\Omega_2} & 0 & A_{43,\Omega_2} & A_{44,\Omega_2} - A_{44,\Omega_4} & -A_{45,\Omega_4} & -A_{46,\Omega_4} \\ 0 & A_{52,\Omega_3} & A_{53,\Omega_3} & -A_{54,\Omega_4} & A_{55,\Omega_3} - A_{55,\Omega_4} & -A_{56,\Omega_4} \\ 0 & 0 & 0 & A_{64,\Omega_4} & A_{65,\Omega_4} & A_{66,\Omega_4} \end{pmatrix},$$

where \mathbf{A} stands for the operator \mathbf{S} , \mathbf{D} , \mathbf{D}^* or \mathbf{N} . Note that the last line of \mathbf{A} must not be computed when $\mathbf{A} = \mathbf{S}$ or \mathbf{D} , nor its last column when $\mathbf{A} = \mathbf{S}$ or \mathbf{D}^* . Indeed, equation (15) is not necessary at the outermost interface and (3b) at the outermost interface directly gives $\vec{n} \cdot \bar{\sigma}_4 \nabla V_{46} = 0$.

The right hand side $\mathbf{b} = [\mathbf{b}_{\gamma_0} \mathbf{b}_{\gamma_1}]^T$ is defined using $\mathbf{b}_{\gamma_p} = [\mathbf{b}_{\gamma_p 1} \mathbf{b}_{\gamma_p 2}, \dots, \mathbf{b}_{\gamma_p N}]^T$ with $p = 0$ or 1 , $N = 5$ if $p = 0$ or $N = 6$ if $p = 1$, and

$$[\mathbf{b}_{\gamma_p \mathbf{i}}]_k = \int_{\mu_k} f_{t_k}(r') b_{\gamma_p \mathbf{i}}(r') dr', \quad (22)$$

where $b_{\gamma_p \mathbf{i}}(r) = -(\gamma_{p_i}^- v_{di p_k}(r) - \gamma_{p_i}^+ v_{di p_l}(r))$, and $f_{t_k} = P_{1k}$ if $p = 0$ and $f_{t_k} = P_{0k}$ if $p = 1$. In the same fashion, the vector \mathbf{a} that contains the coefficients in the unknown expansion reads $\mathbf{a} = [\mathbf{a}_0 \mathbf{a}_1]^T$ with $\mathbf{a}_p = [\mathbf{a}_{p1}, \mathbf{a}_{p2}, \dots, \mathbf{a}_{pN}]^T$ with $p = 0$ or 1 , $N = 5$ if $p = 0$ or $N = 6$ if $p = 1$, and

$$\begin{cases} [\mathbf{a}_{0j}]_l = \beta_{j,l} \\ [\mathbf{a}_{1j}]_l = \alpha_{j,l} \end{cases} \quad (23)$$

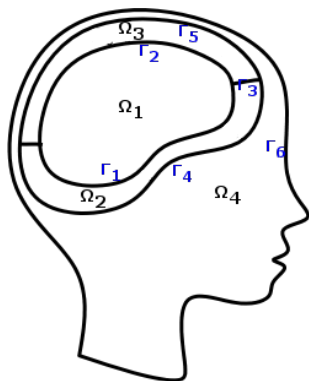


Figure 3: Topology of the domains used to write the exemplificatory system matrix in Section 4.

4.1. Computation of the innermost integral

In practice, computing the elements $[\mathbf{A}_i]_{kl}$ requires two integrations. The inner integral requires to compute either the integral of G_j or of its derivative times the expansion function f_{el} . An analytical solution to obtain this integration for the usual Green's function $G(r, r') = \frac{1}{4\pi} \frac{1}{\|r-r'\|}$ has been proposed in [47]. In (4), the change of variable $R = \sqrt{\bar{\sigma}_i^{-1}} r$ ($\bar{\sigma}_i$ is a symmetric positive definite tensor) transforms $G_i(r, r')$ into $G_i(R, R') = \frac{1}{4\pi} \frac{1}{\|R-R'\|}$ for which we can apply the analytical integration formulas in [47]. The outer integrals (arising from the testing of the operators) are performed numerically using Gaussian integration rules.

4.2. Computation of the hypersingular operator

Using the following representation [42]

$$\begin{aligned}
 & \int_{\mu_{P_{1m}}} P_{1m}(r) N_{ij, \Omega_k} P_{1n}(r) dr \\
 &= \int_{\mu_{P_{1m}} \times \mu_{P_{1n}}} G_k(r, r') \left(\bar{\sigma}_k^{=1/2} \nabla P_{1m}(r) \times \bar{\sigma}_k^{=1/2} \vec{n} \right) \cdot \left(\bar{\sigma}_k^{=1/2} \nabla P_{1n}(r') \times \bar{\sigma}_k^{=1/2} \vec{n}' \right) dr dr' \\
 &= \int_{\mu_{P_{1m}} \times \mu_{P_{1n}}} G_k(r, r') \det(\bar{\sigma}_k) (\bar{\sigma}_k^{-1} \nabla P_{1m}(r) \times \vec{n}) \cdot (\nabla P_{1n}(r') \times \vec{n}') dr dr',
 \end{aligned} \tag{24}$$

$\mathbf{N}_{ij, \Omega_k}$ can be implemented using the computation of $\mathbf{S}_{ij, \Omega_k}$.

4.3. Application of the proposed method to volume meshes

A natural extension of the proposed method is to apply it to volume meshes. In this case, each tetrahedron can be considered as a unique domain with a given conductivity tensor. In regions where the conductivity is isotropic, tetrahedrons with equal conductivities can be merged. This latter property, following from the fact that the formulation we are proposing is a surfacic one in nature, will warranty substantial savings in terms of the number of unknowns with respect to a standard volumetric method. The surface of the isotropic and homogeneous region is then given by the triangles separating tetrahedron with different conductivity tensors. In practice, if a volume mesh is considered, this requires an additional preprocessing step to extract the interfaces between regions with different conductivity tensors so that the considered geometry is a surfacic mesh. This can be done by comparing for each tetrahedron its conductivity tensor with the conductivity tensor of its neighbor. The two tetrahedrons are merged if the difference between their conductivity tensor is smaller than a chosen threshold ϵ . For example, consider the two tetrahedrons T_i and T_j with conductivity tensors $\bar{\sigma}_i$ and $\bar{\sigma}_j$ respectively, and boundaries given by the set of triangles $\bigcup_k t_{ik}$ and $\bigcup_k t_{jk}$ respectively. The interface $t_{ij} = \Gamma_{ij}$ between T_i and T_j is taken into account only if $\|\bar{\sigma}_i - \bar{\sigma}_j\| > \epsilon$, otherwise $T_i \cup T_j$ will form a unique domain Ω_i such that $\partial\Omega_i = \{\bigcup_k t_{ik} \cup \bigcup_k t_{jk}\} \setminus t_{ij}$. To treat the full volume mesh, the process is repeated for all the tetrahedrons. In the case where the mesh is fully inhomogeneous or when no tetrahedron at all are merged, the system matrix \mathbf{Z} will be extremely sparse since for one line, only seven columns of the matrix are not zero. The reader should keep in mind that using volume mesh is not mandatory to use the presented formulation. It is only a convenient pre-processing step in our situation since, moreover, comparison with volume methods (FEM) will be carried out. Indeed, this step is independent from the integral formulation itself and a pure surface mesh can directly be employed. However, the correct modelling of anisotropic compartments requires to subdivide them. As a consequence, while isotropic compartments are modelled with surfaces, anisotropic ones are virtually using volume elements.

5. Numerical Results

Two sets of numerical experiments have been carried out. Since in case of isotropic conductivity, the presented method is the same as the standard symmetric formulation,

we focus on anisotropic conductivity profiles. The first set aimed at assessing the accuracy of the proposed method with canonical anisotropic head meshes (layered anisotropic spheres) while the second set aimed at showing its ability to handle realistic anisotropic conductivity profiles and meshes.

5.1. Modelling of the anisotropy with volume meshes

In modelling the skull layer different approaches are found in the literature. The work of [48] has shown that the different conductivities of the skull layers, a cancellous bone in between two layers of compact bone, are different. As such, it would be recommended, as advise [49], to model these three layers with different conductivities values. However, if this information is not available the skull can also be modelled with one anisotropic layer [49]. Obtaining a precise mesh for the three layers is not without computational overload and it might explains why modeling the skull with a single anisotropic layer is still very common as for example in [27], [28] or [11]. In this paragraph we present results on spherical layered meshes where the skull is modelled with a single anisotropic layer. The method, however, can of course be applied also in the case of simulations where the skull layer is subdivided into its three isotropic composing layers. The first experiments are based on a three layers sphere mesh of radii 0.87, 0.92 and 1. These layers stand for the brain, skull and scalp respectively. In the first and last layer, the conductivity is set to 1 while the conductivity in the middle layer is assumed to be radially anisotropic. The conductivity ratio between the first and the second layer is $\sigma_{skull,iso} = 1/15$, following [50]. The ratio between the radial and tangential conductivity in the skull is 1/10. The radial and tangential conductivity tensor component in each tetrahedra is computed using [27]

$$\frac{4}{3}\pi\sigma_r(\sigma_t)^2 = \frac{4}{3}\pi\sigma_{skull,iso}^3. \quad (25)$$

We begin with a volume mesh and extract a surface mesh by comparing the conductivity values in each tetrahedron following the process described in paragraph 4.3. In this fashion, the tetrahedra in the brain and scalp layers are gathered to form only two homogeneous and isotropic domains while the tetrahedra of the skull layer form different domains depending on the accepted conductivity error ϵ between two neighboring tetrahedra. Figure 4 shows the volume mesh and the surface mesh obtained.

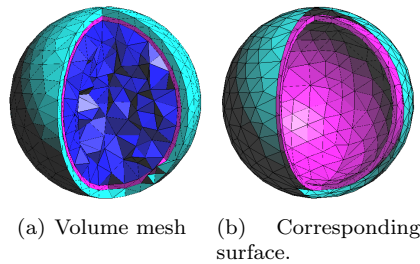


Figure 4: Volume mesh before (a) and after preprocessing, according to the conductivity value of the tetrahedra. (b)

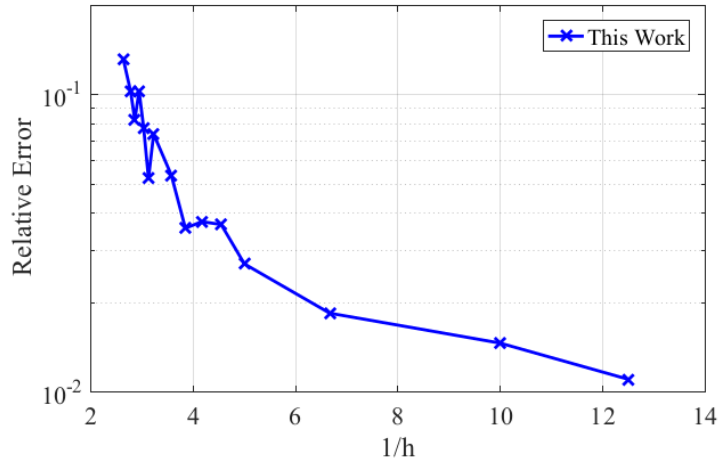


Figure 5: Convergence of the solution of the proposed equation when increasing the number of unknowns.

5.1.1. Convergence of the solution

In this first sets of experiment, ϵ was chosen small enough so that no tetrahedron of the skull layer is merged. This first experiment focused on verifying the convergence of the newly proposed formulation when the number of unknowns is increased. The accuracy is assessed by calculating the relative error of the solution given by the proposed integral formulation with respect to the analytic reference solution available in [13]. The mesh refinement parameter, denoted with h , was decreased from 0.38 to 0.08 so that the initial number of tetrahedra grew from 3607 to 112827. Applying the preprocessing described in 4.3 gave a number of domain increasing from 1495 to 33117. A single dipole source with unitary moment along the z-axis was placed in the center of the innermost sphere to avoid the dipole position influence on the numerical results (studied in a second scenario). The results for this test are shown in Figure 5. It is clear that the relative error decreases, showing the convergence and the consistency of the proposed approach. The time that our prototype code needed to solve the anisotropic EEG forward in the presented canonical scenario is presented Table 1.

Time Z	Time sol.	N elements	sparsity	Relative Error
545 s	12.7 s	19,392	0.1	0.013

Table 1: Timing results to obtain an error of 1% with respect to the analytical anisotropic solution. "Time Z" is the time to build the system matrix, "Time sol" is the time to solve the system (direct inversion), "N elements" is the number of basis functions, sparsity is the ratio of the number of non-zero of the matrix with respect to the size of the matrix.

5.1.2. Accuracy for randomized dipole positions and different dipole eccentricities

Since one dipole may not be representative of the complex brain electric sources pattern of activation, we carried out another experiment, with the same anisotropic conductivity parameters, in which dipole positions are randomly set. For each eccentricity, we computed the relative error in 200 simulations of 1 dipole source, for radial and

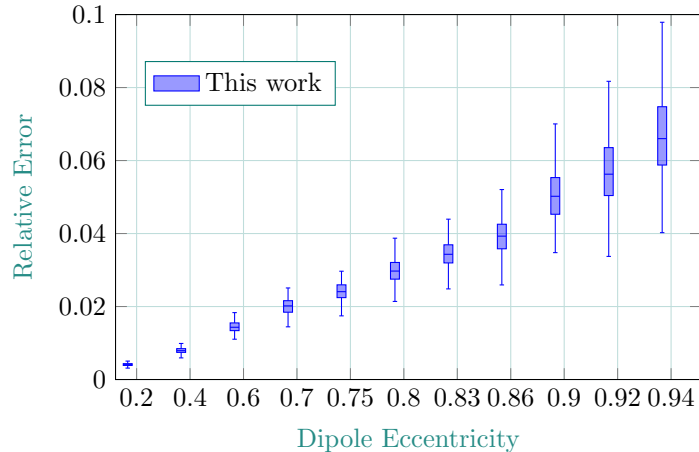


Figure 6: Relative error versus dipole eccentricity in the computation of 200 dipole source both for radial and tangential dipole orientation for the method proposed in this work.

tangential unitary dipole moment. The obtained results, shown in Figure 6, confirm that the proposed approach provides a formulation which is able to handle anisotropic conductivity profile for any dipole position.

5.2. Application to a realistic head mesh

The second experiment is the comparison of the potential obtained in a realistic case with the proposed method and the FEM formulation. Indeed, from grey scale data given by MRI, it is possible to extract volume meshes, (see for example [51]), that are usually used in FEM simulations. From this volume mesh, our method can extract surfaces that delimit domains with the same conductivity. In the presented example, we used [52] to generate a three layers volume head mesh. The fully tetrahedral mesh was used in the FEM solution. A conductivity tensor was associated to each tetrahedra following the choice of the previous anisotropic canonical example. Then, the preprocessing subroutine of our scheme provided a surface mesh for which we applied directly the proposed method. This mesh is shown figure 7(b). The number of tetrahedra was reduced from 187317 to 51709 triangular cells. The choice of a merging threshold $\epsilon = 0.001$ led to the generation of 11698 domains. The obtained scalp potential is shown in figure 8. A relative error of 4.5% between the potential computed with the proposed formulation and the fully anisotropic FEM solution was obtained. The potential computed at 256 electrodes positions is presented in figure 9. This figure shows the overlapping curves obtained with the two methods and confirms the validity of the approach. This realistic scenario shows the ability of the proposed method to handle realistic meshes and in particular, if wanted, the skull anisotropy. In case white matter fiber anisotropy is included, no substantial changes must be done : no assumption was made on the geometry of the media. As a consequence, the only difference would reside in the obtention of the mesh. In practice, the number of degrees of freedom may require the use of a fast solver.

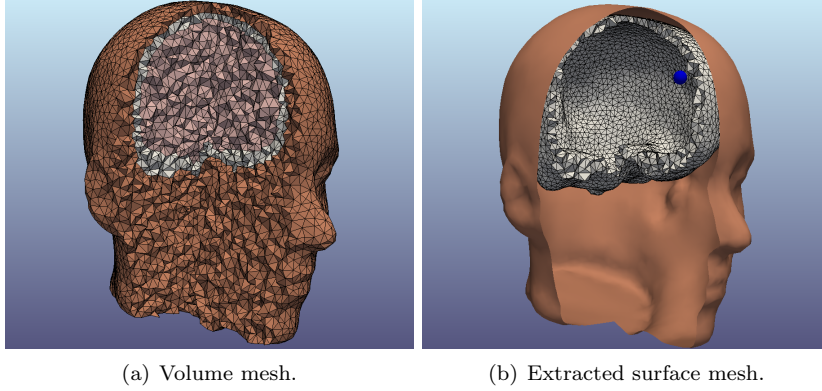


Figure 7: Three Layers mesh used in the realistic simulation : a volume mesh (a) used in the FEM solution and the extracted surface mesh used with the proposed formulation (b). The blue sphere on the right picture shows the position of the dipole used in the simulation.

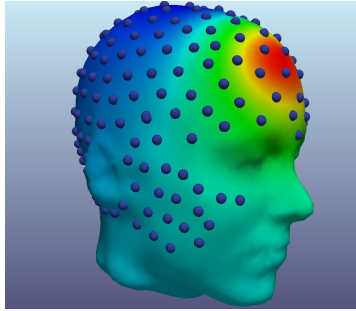


Figure 8: Computed potential on the scalp of a realistic mesh.

6. Discussion and Conclusion

In this paper, a new integral formulation for the EEG forward problem that takes into account anisotropic conductivity profiles was presented. The integral formulation is obtained by leveraging on the representation theorem using an anisotropic fundamental solution. Implementations details are provided and explain the discretization strategy. In particular, a strategy is proposed to tackle inhomogeneous and anisotropic conductivity profile with volume meshes. In this case, the level of inhomogeneity can be controlled by choosing an accepted error in the anisotropic conductivity when building a surfacic mesh from a volume mesh. The level of inhomogeneity does not decrease the accuracy of the solution, on the contrary, it helps to provide a better anatomical model. The numerical results show the correctness of the approach. In the canonical cases, the results given by the new integral equation are matching the analytical solution. In the realistic case, the solution obtained with the proposed formulation is compatible with the one provided by the Finite Elements formulation. One shortcoming of boundary integral formulations, when compared to FEM, is that the arising system matrix is fully populated while in the case of FEM the system matrix is sparse. However, in the presented formulation,

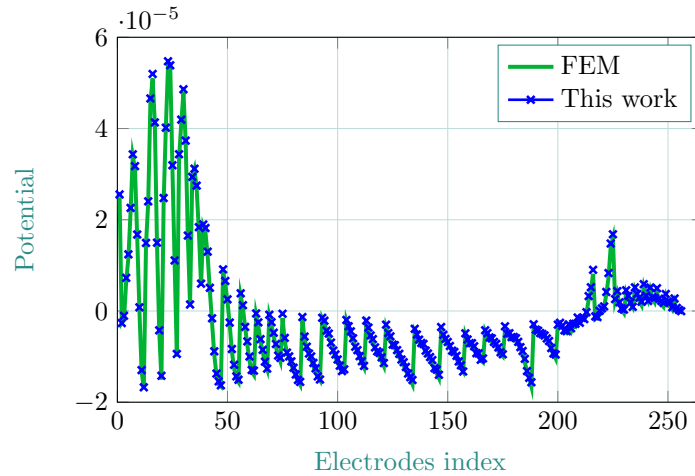


Figure 9: Potential computed at 256 electrodes positions.

as in the symmetric formulation, the number of entries of the system matrix depends on the number of interfaces : the system matrix is block diagonal with the size of the blocks depending on the number of elements in the considered interface. Furthermore, a popular way to fix this disadvantage is to resort to fast solving techniques such as Adaptive Cross Approximation (ACA) or Fast Multipole Method (FMM) that allows to build the system matrix in a linear complexity.

Acknowledgement

This work was supported in part by the CominLabs excellence laboratory, managed by the National Research Agency in the Investing for the Future program under reference ANR-10-LABX-07-01, project SABRE and in part by the European Research Council (ERC) under the European Unions Horizon 2020 research and innovation program (ERC project 321, grant No. 724846).

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