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Characterization of 3-strap antennas in ASDEX Upgrade

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Abstract. The response of the local RF current measured at limiters of 3-strap ICRF antenna to variations of power balance and phasing at $f_{\text{ICRF}}=30\text{MHz}$ agrees qualitatively well with EM calculations by TOPICA and RPLICASOL codes. Measurements of tungsten sputtering yield and DC current at the limiters correlate strongly with the local RF current. In contrast to findings for the 2-strap antennas, values of DC current are predominantly positive, and negative only for some locations and feeding parameters. Explanations can involve more physical mechanisms than only parallel sheath dynamics.

Introduction

Tungsten (W) production due to Ion Cyclotron Range of Frequencies (ICRF) power was drastically reduced by installation and proper feeding of the so-called 3-strap antennas in ASDEX Upgrade (AUG) [1]. Control of the RF image current pattern at the antenna limiters, monitored by local RF probe measurements, was crucial to minimize the ICRF-specific W sources. The measurements were also found to be in a qualitative agreement with electromagnetic (EM) calculations of local near-field at the antenna limiters by TOPICA [2]. This confirmed strong evanescence of the slow wave in the typical AUG conditions of the far scrape-off layer plasma [3]. It also proved the relevance of the approach to optimize the antenna design by lowering local RF currents and RF electrical field at the antenna limiters [4]. Recent experiments in Alcator C-Mod [5] have confirmed that conditions with reduced RF image currents correspond to reduced impurity release and reduced RF-induced plasma potentials.

In this work, we extend the characterization of local quantities in the conditions close to the RF image current cancellation at the limiters of the 3-strap antennas in AUG [1,2], both experimentally (section 1) and theoretically (section 2).

1 Experimental characterization

In [2], experimental characteristics were studied for the case of $f_{\text{ICRF}}=36.5\text{MHz}$. Here we discuss the responses of measurements to variations of strap phasing and power

balance of the 3-strap antenna for $f_{\text{ICRF}}=30\text{MHz}$. This frequency is more convenient for EM modelling, as explained in section 3.

The experimental conditions are similar to those described in [2]: ELM mitigated H-mode plasma at high density, with $P_{\text{ICRF}}=1\text{MW}$ with H-minority heating and $P_{\text{NBI}}=5\text{MW}$, but at $B_t=2.0\text{T}$ in discharge #33105. Figure 1 shows the 3-strap antenna 4 (*a4*) in sector 12 of AUG and the locations of the following measurements: RF current (expressed as 50 Ohm equivalent voltage V_{RF}), DC current I_{DC} at limiter tiles numbered 01 to 12, as well as spectroscopic measurements of W sputtering yield Y_{W} close to limiter tiles 01 to 03 and 07 to 09. Figures 2 and 3 presents V_{RF} and I_{DC} expressed as 2D functions of the strap power balance $P_{\text{central}}/P_{\text{outer}}$ (where P_{central} is the power to central strap and P_{outer} is the sum of the power to the outer straps) and of the deviation of phasing between the central strap and the in-phase outer straps from dipole $\Delta\Phi$, for left and right antenna sides, respectively. Figure 4 shows Y_{W} .

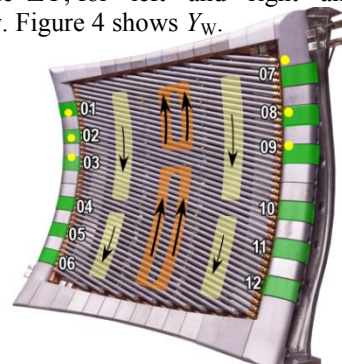


Fig. 1. 3-strap antenna with strap RF currents and measurement locations highlighted.

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** See H. Meyer et. al. Nuclear Fusion FEC 2016 Special issue (2017)

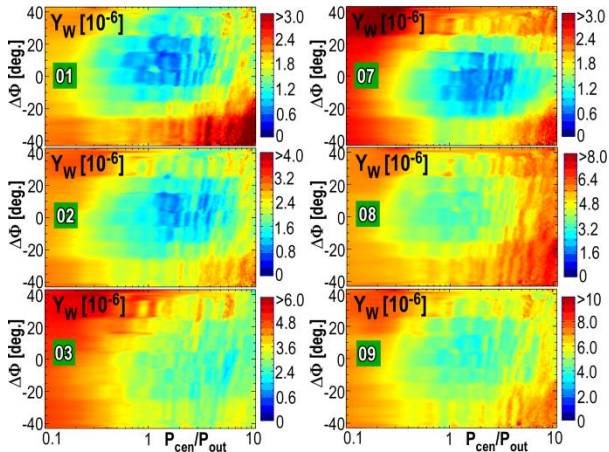


Fig. 4. Y_W as a function of $P_{\text{central}}/P_{\text{outer}}$ and $\Delta\Phi$ at locations 01, 02, 03 at the left and 07, 08, 09 at the right antenna side (note different colour scales).

2 Theoretical characterization

The case with $f_{\text{ICRF}}=30\text{MHz}$ is better suited for modelling with EM codes, because the frequency is significantly lower than the resonant frequency of the straps, in particular that of the outer ones. Depending on model and plasma conditions, the latter is in the range between 36.5 and 36.9MHz. Crossing this resonance frequency does not affect the near-field distribution significantly. However working at $f_{\text{ICRF}}=30\text{ MHz}$ helps to resolve issues in calculations with the RAPLICASOL [10] code used in this work.

We consider three cases modeled with EM codes, all using the same set of plasma profiles corresponding to the scenario of discharges #31515 and #33105. The first one uses a flat model of the 3-strap antenna in TOPICA as in [2]. The second one introduces a more realistic model of the antenna in curved geometry in TOPICA [11]. The third case applies the same flat geometry as in the first case, but in RAPLICASOL code.

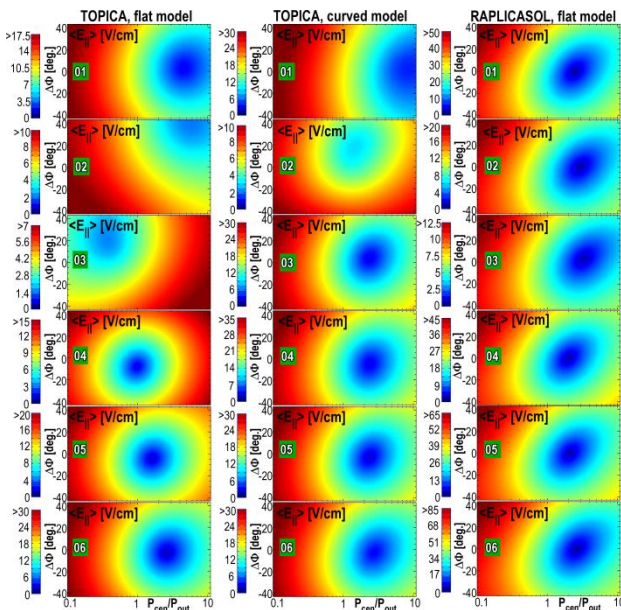


Fig. 5. $\langle E_{\parallel} \rangle$ calculated with TOPICA flat (left), TOPICA curved (middle) and RAPLICASOL flat (right) models at the left antenna side locations (note different colour scales).

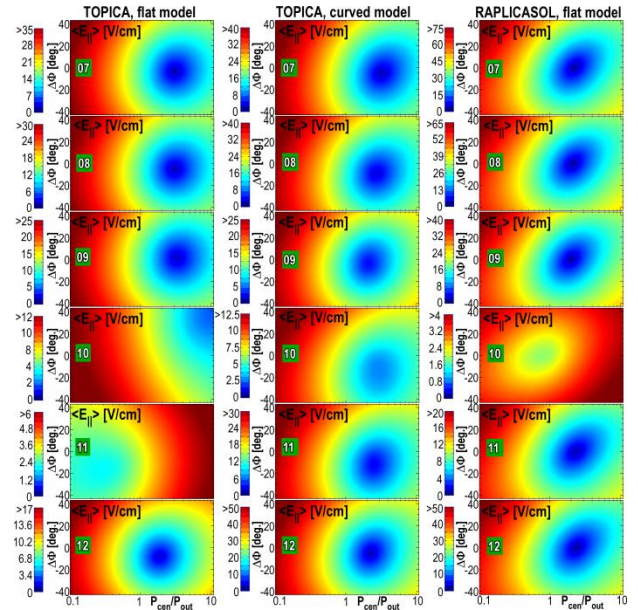


Fig. 6. $\langle E_{\parallel} \rangle$ calculated using TOPICA flat (left), TOPICA curved (middle) and RAPLICASOL flat (right) models at the right antenna side locations (note different colour scales).

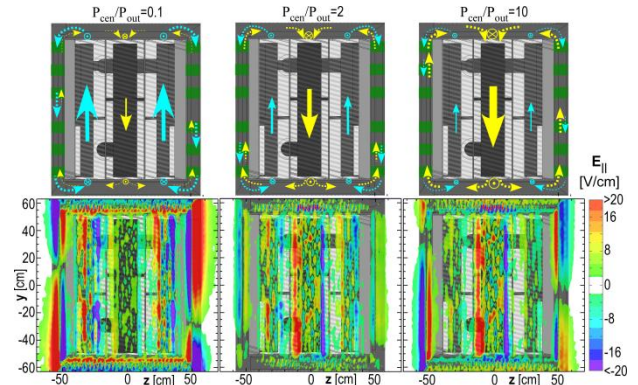


Fig. 7. E_{\parallel} for flat TOPICA model for different cases of power balance in dipole for 30 MHz, with graphics showing RF current circulation at limiters (top). Purely vertical dashed arrows on the limiters indicate locations of sign reversal of parallel component of RF current.

As a theoretical “proxy”, we use spatially averaged values of the E_{\parallel} field $\langle E_{\parallel} \rangle$ calculated just in front of the antenna limiters at locations 01-12. Firstly, due to strong evanescence of the slow wave, $\langle E_{\parallel} \rangle$ is the main contributor to the RF sheath driving voltage and rectified sheath voltage, as has been discussed in [2] and has been shown theoretically in [3,12]. Secondly, this quantity is primarily a linear function of local RF current (in particular, of its parallel component) at fixed geometrical and loading conditions. Thirdly, $\langle E_{\parallel} \rangle$ is an engineering quantity which can easily be used for antenna optimization with practically any EM code. The ICRF power is scaled to 500 kW for all the simulations.

Figures 5 and 6 show the $\langle E_{\parallel} \rangle$ calculations in the three different modelling cases (columns) for the left and the right antenna sides, respectively. Location-dependent areas with minima close to the dipole phasing and to $P_{\text{central}}/P_{\text{outer}} \approx 2$ are well reproduced in all three modelling cases compared to the experimental V_{RF} diagrams, with details which differ. These minima correspond to the conditions of the local cancellation of all components of

RF current. Spatial E_{\parallel} distribution is shown in Fig. 7 for $P_{\text{central}}/P_{\text{outer}}$ of 0.1, 2.0 and 10.0 in dipole phasing calculated for flat model in TOPICA, together with qualitative presentation of the RF image current circulation on the limiters.

According to Fig.5 and Fig.6, minima of $\langle E_{\parallel} \rangle$ become broader somewhat above the midplane for the left side (locations 02, 03) and somewhat below the midplane on the right side (locations 10, 11). In experiment, broader regions with minima of V_{RF} are also observed on the right side at about the midplane (locations 09, 10 in Fig. 3). Close to the regions with broader minima, $\langle E_{\parallel} \rangle$ are generally lower, according to the calculations. Here, parallel projection of the RF image current of individual straps, flowing on the limiters, changes sign (unbalanced cases in Fig.7), as well as the total RF current at the limiters is reduced. This results in minima at the E_{\parallel} maps. However in the experiment, there is no clear tendency for significantly lower values in the locations corresponding to these.

The more detailed model with curved geometry in TOPICA produces coupling parameters which match experimental values significantly better [11]. Consequently, a more realistic relation between the power balance and the voltage balance inside the antenna (presented in [1]), is reproduced and the confidence in results using the curved antenna model is higher.

Although the general trends of the local quantities are well described by the simulations, and one can speak of at least a qualitative agreement; quantitative one-to-one correspondence between the experiments and the calculations is difficult due to existing limitations and uncertainties. As can be inferred from [6], details of 3D density profile have a significant effect on the RF image currents and E_{\parallel} at the limiters. On top of that, the profiles are not stationary in experiments [2]. Even if a better qualitative correspondence were found by refining experimental input and by detailing the models further, the variety of the profiles in the experiment would make it very challenging to optimize antenna design further, in order to have minima at the same values of $P_{\text{central}}/P_{\text{outer}}$ and $\Delta\Phi$ at all limiter locations, and this for a broad variety of plasma conditions.

One of alternative approaches to optimize the 3-strap antenna further in experiments is increasing the number of actuators by making the antenna limiters active, fed independently by feedback-controlled RF sources. This could allow counteracting the residual RF currents at the limiters by destructive interference. However this approach needs testing on small-scale machines. Another actuator which already is available in the experiments and planned to be used in the future to tackle RF fields in the scrape-off layer is the phasing between antenna pairs.

Summary

Experimental local values of RF current, DC current, W sputtering yield were presented for multiple locations at the 3-strap antenna as functions of variations of antenna strap power balance and strap phasing. The quantities experience a minimum close to dipole phasing when

power of the central strap is close to the double of the sum of the power to the outer straps.

The DC current can however reverse sign and is predominantly positive in contrast to previous studies with 2-strap antennas [7]. This implies a complicated nature of the DC circuit, which can also involve a collection of ion current transverse to the magnetic field.

Comparison of measured RF current at the limiters with EM calculations using a flat antenna model in TOPICA and RPLICASOL as well as a more realistic curved antenna model in TOPICA shows that qualitative features of the experimental observations are captured by the calculations. Better quantitative agreement between simulations and measurements is presenting a challenging task, given the uncertainties and variations in the experimental input [2,6]. An independent method to reduce the RF currents by actively feeding the antenna limiters and synchronizing operation of all four antennas in ASDEX Upgrade could be tested in the future.

We refer to [13] where the recent progress in solving the problems of the antenna-plasma interface in ASDEX Upgrade is summarized.

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