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SNR Gap Between MIMO Linear Receivers: Characterization and Applications

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Abstract—This paper presents a statistical characterization of the SNR gap between MIMO Zero-Forcing (ZF) and Minimum Mean Squared Error (MMSE) equalizers, beyond the Rayleigh assumption for the interfering streams amplitude fading. Results are valid for arbitrary transmit SNR values and number of transmit/receive antennas. Specifically, we provide the exact closed-form distribution of the random variable representing the difference between the output SNR on a generic receive filter branch, under MMSE and ZF equalization. Analytical results turn particularly useful for the study of heterogeneous cellular networks.

I. INTRODUCTION

Zero Forcing (ZF) and Minimum Mean Squared Error (MMSE) are the most popular linear receivers, due to their satisfactory trade-off between implementation complexity and achievable performance. Their supposed performance equivalence in the high Signal to Noise Ratio (SNR) regime was contradicted in [1]. Therein, it was shown that, given the transmission of n_t signal streams, with common value of the normalized transmit SNR, the output SNR for independent stream decoding, γ_k^{mmse} , measured in correspondence of an arbitrary branch of the MMSE receiving filter, can be expressed as [1], [2]

$$\gamma_k^{\text{mmse}} = \gamma_k^{\text{zf}} + \eta_k \quad (1)$$

In (1), γ_k^{zf} represents the output SNR, of the ZF equalizer on branch k and η_k denotes a non-decreasing function of the SNR, accounting for the energy nulled out by the ZF but not by the MMSE receiver. A gap between the output SNR values strongly impacts on both the outage (see [1, Eq. (53)]) as well as the error probability (EP), as outlined, e.g., in [1, Eq. (50)]. Moreover, knowledge of the statistics of η_k can serve as a high-SNR upper bound to the Interference to Noise Ratio at the output of an MMSE filter [1, Lemma III.2]. Due to its relevance as a performance index for linear receivers, hereinafter we investigate the statistics of η_k .

Related work. To date, the probability density function (pdf) of η_k has been derived in closed form only for asymptotically large transmit power. In [1, Thm. III.1], it was shown that the pdf of a properly scaled η_k , as the SNR grows large, converges to the $\mathcal{F}_{2(n_t-2), 2(n_r-n_t+2)}$ distribution¹, under the assumption that both the intended, as well as the interfering streams, are

subject to uncorrelated Rayleigh fading. The case of a system with a large number of transmit and receive antennas, impaired by transmit-correlated Rayleigh fading, has been investigated in [2]. Irrespectively of the presence of spatial correlation, under Rayleigh fading, γ_k^{zf} is Gamma distributed [3].

Very recent works [4, and references therein] deal with the distribution of γ_k^{zf} in the Line-of-Sight (LOS) environment, i.e., where either the desired or the interfering streams [5] experience Rician fading, while remaining streams experience Rayleigh fading. No results on either γ_k^{zf} or η_k are available for other channel models. The exact characterization of γ_k^{mmse} , in turn, is a more challenging task², calling for the separate investigation of γ_k^{zf} and η_k , which are easier to handle.

Aim of the work. We capitalize on the fact that η_k , in [1, Formula (26)], is cast in terms of an Hermitian quadratic form, where the desired signal stream only appears in the vector, while the interfering streams are confined to the kernel matrix. This algebraic structure strongly favors its explicit statistical characterization, which can be carried out by means of standard analytical tools of multivariate analysis in several fading cases.

We focus on the case of Rayleigh-faded intended stream. Under this assumption, we first provide a non-asymptotic result complementing the fundamental one of [1, Thm. III.1], by giving the gap distribution for Rayleigh-faded interfering streams but arbitrary rather than arbitrarily large SNR values. The case of transmit-correlated Rayleigh fading, possibly corresponding to spatially-clustered interferers, is addressed in the same vein. Then, we derive the corresponding statistics in the case of Rician-faded interfering streams, for arbitrary rank and eigenvalues multiplicity of the LOS matrix. Such a scenario is representative of a worst-case cellular transmission, where the useful stream comes, e.g., from the cell edge and dominant interferers have LOS path toward the base station. Finally, we evaluate the pdf of η_k for the case where interfering streams undergo multiple Rayleigh scattering³. This last scenario is of particular interest in foreseen small-cells networks. Indeed, it

²To the best of the authors' knowledge, the pdf of γ_k^{mmse} is available when both desired and interfering signals are Rayleigh faded [6], while a closed-form expression for its moments has been derived also in the case of Rice-faded intended stream/Rayleigh faded interference [7].

³See [10] for the statistical analysis of a multiple scattering system, and [11], [12], [13, and references therein] for practical justification and information-theoretic performance of such channels, as the number of scattering stages vary.

¹I.e., the pdf of $\mathcal{F}_{m,n}$ is: $f_x(x) = \frac{\Gamma(\frac{n+m}{2})n^{n/2}m^{m/2}x^{m/2-1}}{\Gamma(m/2)\Gamma(n/2)(n+mx)^{\frac{n+m}{2}}}$.

adequately represents, with no approximation in the channel statistics, the case where interfering signals come from non colocated single-antenna equipped users, whose signals may undergo multiple scattering phenomena.

As an instance of application of our results, we evaluate the key statistics of η_k , relating the values of the Bit Error Probability (BER) as well as those of the Outage Probability (OP) of an MMSE receiver with the corresponding performance indices in the case of ZF.

II. NOTATION

Boldface uppercase and lowercase letters denote matrices and vectors, respectively. \mathbf{I} is the identity matrix. The determinant and the conjugate transpose of the generic matrix \mathbf{A} are denoted by $|\mathbf{A}|$ and \mathbf{A}^H , respectively, while $\mathbf{A}_{i,j}$ is the (i, j) -th element of \mathbf{A} . Moreover, $\mathbb{E}_a[\cdot]$ represents the average operator with respect to the random variable a . For any $m \times m$ Hermitian matrix \mathbf{A} with eigenvalues a_1, \dots, a_m , the Vandermonde determinant is defined as: $\mathcal{V}(\mathbf{A}) = \prod_{1 \leq k < \ell \leq m} (a_k - a_\ell)$. $G_{a,b}^{c,d}(\cdot)$, with integer parameters a, b, c, d , denotes the Meijer-G function [14, Ch. 8]. The Gauss hypergeometric function is denoted by ${}_2F_1(a, b; c; x)$. The complex multivariate Gamma function is defined as [15]: $\Gamma_p(q) = \pi_p \prod_{\ell=1}^p (q - \ell)!$, with p and q non-negative integers such that $p \leq q$, and $\pi_p = \pi^{p(p-1)/2}$. We also denote by $(a)_k = \Gamma(a+k)/\Gamma(a)$ the Pochhammer symbol of order k . $f_a(a)$ denotes the pdf of the scalar random variable a (for random matrices pdf we skip the subscript). If two variables a and b share the same distribution, we write $a \sim b$.

III. SYSTEM MODEL

Consider a linear system:

$$\mathbf{y} = \sqrt{\rho} \mathbf{H} \mathbf{x} + \mathbf{n} \quad (2)$$

where \mathbf{y} is a vector of size n_r , \mathbf{H} is the $n_r \times n_t$ random channel matrix, \mathbf{x} is a random vector of size n_t with covariance $\mathbb{E}[\mathbf{x}\mathbf{x}^H] = \mathcal{E}_s/n_t \mathbf{I}$, and \mathbf{n} represents Gaussian noise with covariance $\mathbb{E}[\mathbf{n}\mathbf{n}^H] = \mathcal{N}_0 \mathbf{I}$. For simplicity, we assume $n_t \leq n_r$. The channel matrix is assumed to be normalized such that $\rho = n_r n_t / \text{Tr}\{\mathbb{E}[\mathbf{H}\mathbf{H}^H]\}$.

The expression of the instantaneous received SNR on the k -th receive filter branch ($k = 1, \dots, n_t$) depends on the adopted equalizer. In terms of the channel matrix and system parameters, we have [16, Ch. 6]:

$$\gamma_k^{\text{mmse}} = \frac{1}{\left[(\mathbf{I} + \delta \mathbf{H}^H \mathbf{H})^{-1} \right]_{k,k}} - 1, \quad \gamma_k^{\text{zf}} = \frac{\delta}{\left[(\mathbf{H}^H \mathbf{H})^{-1} \right]_{k,k}}, \quad (3)$$

where $\delta = \frac{\mathcal{E}_s}{\mathcal{N}_0} \frac{\rho}{n_t}$ and [1, Eq. (26)]. We denote by \mathbf{h}_k the k -th column of \mathbf{H} and by \mathbf{H}_k the matrix obtained by removing \mathbf{h}_k from \mathbf{H} . The SNR gap $\gamma_k^{\text{mmse}} - \gamma_k^{\text{zf}}$ is then given by [1]

$$\eta_k = \mathbf{h}_k^H \mathbf{U}_k (\delta^{-1} \mathbf{I}_{n_t-1} + \mathbf{\Lambda}_k)^{-1} \mathbf{U}_k^H \mathbf{h}_k, \quad (4)$$

where $\mathbf{H}_k = \mathbf{U}_k \mathbf{\Lambda}_k^{1/2} \mathbf{V}_k^H$ is the singular value decomposition of \mathbf{H}_k , \mathbf{U}_k has size $n_r \times (n_t - 1)$ and both $\mathbf{\Lambda}_k$ and \mathbf{V}_k are square matrices of size $(n_t - 1)$.

IV. PROBLEM FORMULATION

Our work focuses on the statistical characterization of (4), under different assumptions on the statistics of the diagonal matrix $\mathbf{\Lambda}_k$ of size $n_t - 1$, while $\mathbf{h}_k \sim \mathcal{CN}(0, \mathbf{I}_{n_r})$, i.e. \mathbf{h}_k is a length- n_r vector of zero-mean uncorrelated Gaussian random variable with unit variance. This assumption on the intended stream allows us to state the following lemma on the gap distribution, conditionally on the interfering streams fading law.

Lemma 4.1: For independent stream decoding, the SNR gap η_k , as defined in (4), is a random variable whose conditional law w.r.t. the interfering streams fading distribution can be expressed as

$$f_{\eta_k | \mathbf{\Lambda}_k}(y) = \frac{|\Delta|}{\mathcal{V}(\mathbf{\Lambda}_k)} \prod_{\ell=1}^{n_t-1} \left(\frac{1}{\delta} + \lambda_\ell \right)^{n_t-2}. \quad (5)$$

Δ is the determinant of a size- $(n_t - 1)$ matrix, with generic element

$$\Delta_{i,j} = \begin{cases} \ell_i^{j-1}, & 1 \leq i \leq n_t - 1, 1 \leq j \leq n_t - 2 \\ \ell_i^{n_t-3} e^{-y/\ell_i}, & 1 \leq i \leq n_t - 1, j = n_t - 1 \end{cases}$$

$$\ell_i = \left(\frac{1}{\delta} + \lambda_i \right)^{-1}.$$

Proof: Since $\mathbf{h}_k \sim \mathcal{CN}(0, \mathbf{I}_{n_r})$, $\mathbf{U}_k^H \mathbf{h}_k \sim \mathcal{CN}(0, \mathbf{I}_{n_t-1})$, by the invariance of the complex Gaussian distribution to linear transformations. Henceforth, the conditional distribution of $\eta_k | \mathbf{\Lambda}_k$ can be written directly as per [17, Eq. (8)]. ■

Equipped with (5), we derive its unconditional expression for each listed fading scenario. For sake of formula's compactness, we define $\zeta = \frac{1+\delta}{\delta}$ and $\tau = n_r - n_t + 1$. Due to the lack of space, the proofs of the following Propositions are omitted, and relegated to [20].

A. Rayleigh fading

Under the uncorrelated Rayleigh assumption for the interferers, \mathbf{H}_k is modeled as a matrix whose entries are zero-mean independent complex Gaussian random variables with unit variance. As δ takes on arbitrary but finite values, η_k 's law can be expressed as per the following proposition.

Proposition 4.1: The distribution of (4), under the assumption of Rayleigh-faded interfering streams, can be written as

$$f_{\eta_k}(y) = \mathcal{K}_{n_t-1, n_r} |\mathbf{Z}|, \quad (6)$$

where

$$\mathcal{K}_{n_t-1, n_r} = \frac{\pi_{n_t-1}^2}{\Gamma_{n_t-1}(n_t-1) \Gamma_{n_t-1}(n_r)},$$

$$\mathbf{Z}_{i,j} = \begin{cases} \beta! \sum_{\ell=0}^{\beta} \frac{(n_t-j-\ell)_{n_r-i}}{\delta^\ell \ell!} & 1 \leq i \leq n_t-1 \\ & 1 \leq j \leq n_t-2 \\ \frac{e^{-y/\delta} (n_r-i)!}{\delta (y+1)^{n_r-i+1}} \left[1 + \frac{\delta (n_r-i+1)}{y+1} \right] & 1 \leq i \leq n_t-1 \\ & j = n_t-1 \end{cases}$$

and, hereinafter, $\beta = n_t - j - 1$.

Remark 1: Notice that (6) can be also cast as

$$f_{\eta_k}(y) = \mathcal{K}_{n_t-1, n_r} \sum_{i=1}^{n_t-1} \frac{\mathcal{D}_i e^{-y/\delta} (n_r-i)!}{\delta (y+1)^{n_r-i+1}} \left[1 + \frac{\delta (n_r-i+1)}{y+1} \right], \quad (7)$$

by virtue of Laplace expansion of $|\mathbf{Z}|$ w.r.t. its last column, the only one depending on η_k . In (7), \mathcal{D}_i is the $(i, n_t - 1)$ -th cofactor of the matrix \mathbf{Z} . Such an expression holds for each of the following newly derived pdfs. Though less compact than that appearing in the Proposition's statements, (7) turns out to be effective in averaging w.r.t. η_k , under any fading scenario.

Corollary 4.1: Under the assumptions of Proposition 4.1, if the rows of \mathbf{H}_k are correlated with common covariance matrix Σ_k , with distinct⁴ eigenvalues σ_i 's, $i = 1, \dots, n_t - 1$

$$f_{\eta_k}(y) = \frac{\pi_{n_t-1} |\mathbf{Z}|}{\Gamma_{n_t-1}(n_r) \mathcal{V}(\Sigma_k) |\Sigma_k|^{\tau+1}}, \quad (8)$$

$$\mathbf{Z}_{i,j} = \begin{cases} \sum_{\ell=0}^{\beta} \beta! \sigma_i^{n_r-j+1} (n_r-j-\ell)! & 1 \leq i \leq n_t-1, \\ (\delta \sigma_i)^\ell \ell! (\beta-\ell)! & 1 \leq j \leq n_t-2 \\ \frac{e^{-y/\delta} \tau!}{\delta \left(\frac{1}{\sigma_i} + y\right)^{\tau+1}} \left[1 + \frac{\delta(\tau+1)}{\left(\frac{1}{\sigma_i} + y\right)} \right] & 1 \leq i \leq n_t-1, \\ & j = n_t-1 \end{cases}$$

B. Rice fading

Assuming that all interfering streams undergo Rice fading is tantamount to say that

$$\mathbf{H}_k = \bar{\mathbf{H}}_k + \tilde{\mathbf{H}}_k. \quad (9)$$

In the above expression $\bar{\mathbf{H}}_k$ is a deterministic matrix, representing the LOS signal component between each interfering transmitter and each receiver antenna. The entries of $\tilde{\mathbf{H}}_k$ are independent, zero-mean complex Gaussian random variables. We provide hereinafter the gap characterization for arbitrary geometric behavior of the LOS component, i.e., we assume the matrix $\Omega = \bar{\mathbf{H}}_k^H \bar{\mathbf{H}}_k$ to have rank $L \leq n_t - 1$. We denote by $\{\omega_1, \dots, \omega_L\}$ its L non-zero eigenvalues.

Proposition 4.2: The distribution of the Hermitian quadratic form in (4), under the assumption of Rice-faded interfering streams, with rank- L LOS matrix, can be written as

$$f_{\eta_k}(y) = \mathcal{K}_\Omega |\mathbf{Z}|, \quad (10)$$

where

$$\mathcal{K}_\Omega = \frac{e^{-\text{Tr}\{\Omega\}} \prod_{\ell=1}^L \omega_\ell^{L+1-n_t}}{\Gamma_{n_t-L-1}(n_t-L-1) \mathcal{V}(\Omega)},$$

$\mathcal{V}(\Omega) = \prod_{\ell < k}^L (\omega_\ell - \omega_k)$ and

$$\mathbf{Z}_{i,j} = \begin{cases} \sum_{\ell=0}^{\beta} \frac{(n_r-j-\ell)!}{e^{-\omega_j} \ell! \delta^\ell} & 1 \leq i \leq L \\ \times \sum_{q=0}^{\beta-\ell} \frac{(n_t-j-\ell-q)_{t+q}}{\omega_j^{-q} q! (\tau+q)!} & 1 \leq j \leq n_t-2 \\ \frac{\beta!}{(n_r-i)!} \sum_{\ell=0}^{\beta} \frac{(n_t-j-\ell)_{n_r-i}}{\ell! \delta^\ell} & L+1 \leq i \leq n_t-1 \\ & 1 \leq j \leq n_t-2 \\ \frac{e^{-y/\delta + \omega_i}}{\delta(1+y)^{\tau+1}} \left[1 + \frac{\delta(\tau+1 + \omega_i)}{y+1} \right] & 1 \leq i \leq L \\ & j = n_t-1 \\ \frac{e^{-y/\delta}}{\delta(y+1)^{n_r-i+1}} \left[1 + \frac{\delta(n_r-i+1)}{1+y} \right] & L+1 \leq i \leq n_t-1 \\ & j = n_t-1 \end{cases}$$

⁴The case of multiple eigenvalues can be addressed resorting to classical limiting procedures, which we do not report here for lack of space.

Corollary 4.2: Under the assumption of Proposition 4.2, if $L = n_t - 1$ and $\omega_i = \alpha n_r$, $i = 1, \dots, n_t - 1$, $\alpha > 0$,

$$f_{\eta_k}(y) = \frac{e^{-\alpha n_r (n_t-1)} |\mathbf{Z}|}{\Gamma_{n_t-1}(n_t-1)}, \quad (11)$$

where

$$\mathbf{Z}_{i,j} = \begin{cases} e^{\alpha n_r} \sum_{\ell=0}^{\beta} \frac{(n_r+n_t-i-j-\ell-1)!}{\ell! \delta^\ell} & \\ \times \sum_{q=0}^{\beta-\ell} \frac{(n_t-j)_{\ell+q} (\alpha n_r)^q}{q! (n_r-i+q)!} & 1 \leq i \leq n_t-1 \\ & 1 \leq j \leq n_t-2 \\ \frac{e^{-y/\delta + \alpha n_r / (1+y)}}{\delta(1+y)^{n_r-i+1}} & \\ \times \left[1 + \frac{\delta(n_r-i+1 + \frac{\alpha n_r}{1+y})}{y+1} \right] & 1 \leq i \leq n_t-1 \\ & j = n_t-1 \end{cases}$$

C. Small-cells and multiple scattering

Consider a multiple-scattering channel with $N - 1$ clusters of n_i independent scatterers each. In this case the matrix \mathbf{H}_k can be represented by the product of N matrices, \mathbf{S}_i , of size $n_i \times n_{i-1}$, $i = 1, \dots, N$, with $n_0 = n_t - 1$ and $n_N = n_r$. The entries of \mathbf{S}_i are zero-mean unit variance complex Gaussian independent random variables. For sake of compactness, we define the set of auxiliary variables $\nu_i = n_i - n_0$, $i = 1, \dots, N$. In the following we assume $n_t - 1 \leq n_1 \leq \dots \leq n_r$, thus such variables are non-negative integers⁵.

Proposition 4.3: The distribution of (4), under the assumption of multiple Rayleigh scattering affecting the interfering streams, can be written as

$$f_{\eta_k}(y) = \frac{|\mathbf{Z}|}{\prod_{i=1}^{n_0} \prod_{\ell=0}^N \Gamma(i + \nu_\ell)} \quad (12)$$

where

$$\mathbf{Z}_{i,j} = \begin{cases} \sum_{\ell=0}^{\beta} \binom{\beta}{\ell} \frac{\Gamma(n_1+i-j-\ell)}{\delta^\ell} & \\ \times \prod_{q=2}^N (n_q-j-\ell)!, & 1 \leq i \leq n_t-1 \\ & 1 \leq j \leq n_t-2 \\ \frac{e^{-y/\delta}}{\delta} \left[G_{N,1}^{1,N} \left(\begin{matrix} -\nu_i \\ 1 \end{matrix} \middle| y \right) \right. & \\ \left. + \delta G_{N,1}^{1,N} \left(\begin{matrix} -\nu_i-1 \\ 1 \end{matrix} \middle| y \right) \right] & 1 \leq i \leq n_t-1 \\ & j = n_t-1 \end{cases}$$

and $\nu_i = [\nu_N, \dots, \nu_2, \nu_1 + i - 1]$.

V. APPLICATIONS

Here we present two main applications of the SNR gap statistics evaluation, focusing on the BER and the OP.

⁵This assumption can be relaxed based on the observations in [10].

A. BER of ZF and MMSE via SNR gap statistics

In presence of a BPSK-modulated input signal, ZF and MMSE uncoded BER are linked as per [1, Eq. (50)]. In a non-asymptotic setting, with reference to M -PSK⁶,

$$P_b^{\text{mmse}} \approx \mathbb{E}_{\eta_k} [e^{-\eta_k \sin^2(\pi/M)}] P_b^{\text{zf}}. \quad (13)$$

With reference to the above equation and setting $M = 2$ for simplicity, we can write the required expectation in the uncorrelated Rayleigh case as

$$\mathbb{E}_{\eta_k} [e^{-\eta_k}] = \mathcal{K}_{n_t-1, n_r} \sum_{i=1}^{n_t-1} \frac{\mathcal{D}_i}{\delta} [\mathcal{I}_{n_r-i, \zeta} + \delta \mathcal{I}_{n_r-i+1, \zeta}], \quad (14)$$

and, in the rank- L Rician case, as

$$\mathbb{E}_{\eta_k} [e^{-\eta_k}] = \mathcal{K}_{\Omega} \left[\sum_{j=1}^L \frac{e^{\mu_j} \mathcal{D}_j}{\tau! \delta} \left(\mathcal{I}_{\tau, \zeta} + \frac{\delta(\tau + \mu_j + 1) \mathcal{I}_{\tau+1, \zeta}}{\tau + 1} \right) \sum_{j=1+L}^{n_t-1} \frac{\mathcal{D}_j}{\delta} (\mathcal{I}_{n_r-j+1, \zeta} + \delta \mathcal{I}_{n_r-j+2, \zeta}) \right], \quad (15)$$

where

$$\mathcal{I}_{m, \zeta} = (-\zeta)^m \left[\sum_{\ell=1}^m \frac{(\ell-1)!}{(-\zeta)^\ell} - e^{\zeta} \text{Ei}(-\zeta) \right].$$

Finally, in presence of multiple scattering,

$$\mathbb{E}_{\eta_k} [e^{-\eta_k}] = \sum_{i=1}^{n_t-1} \frac{\mathcal{D}_i}{\mathcal{Z}_{n_t-1} \delta} \left[G_{1, N+1}^{N+1, 1} \left(\begin{matrix} -1 \\ 0, \bar{\nu}_i \end{matrix} \middle| \zeta \right) + \delta G_{1, N+1}^{N+1, 1} \left(\begin{matrix} -1 \\ 0, \bar{\nu}_i + 1 \end{matrix} \middle| \zeta \right) \right] \quad (16)$$

where $\bar{\nu}_i = [\nu_N, \dots, \nu_1 + i]$.

Remark III: Notice that (14), together with [1, Eq. (39)], provides an analytical approximation of P_b^{mmse} under Rayleigh fading, under the assumptions of Gaussian approximation for the interference and non rank-deficient channel⁷. Numerical results, not reported herein due to lack of space, confirm also, for moderate values of δ , the presence of an offset between ZF and MMSE per-stream BER, as observed in [1, Fig. 3]. Extension to different fading laws and/or signal constellation is subject of ongoing work.

B. Outage Probability evaluation

Employing independent codes of rate R over each of the transmit antennas, the MMSE k -th stream OP can be evaluated from⁸ [1, Eq. (52)], namely

$$P_{\text{out}, k}^{\text{mmse}}(R) = \int_0^{2^R-1} F_{\gamma_k^{\text{zf}}} \left(\frac{2^R-1-x}{\delta} \right) f_{\eta_k}(x) dx, \quad (17)$$

which once again only requires the cumulative distribution of γ_k^{zf} and the density of η_k . In the case of Rayleigh-faded

⁶Our results can be applied to M -QAM constellations, too, but we do not report this case here due to space limitations.

⁷See again [1] for a detailed discussion on those assumptions' validity.

⁸We put in the OP expression the generic CDF of γ_k^{zf} , instead that the one used in [1, Eq. (52)], whose expression only refers to Rayleigh fading.

interfering streams, (17) can be expressed in closed form by substituting into (17) the expression for $f_{\eta_k}(x)$ given in (7) and the expression for $F_{\gamma_k^{\text{zf}}}(x)$ given in [1], i.e., $F_{\gamma_k^{\text{zf}}}(x) = 1 - e^{-x} \sum_{\ell=0}^{n_r-n_t} \frac{x^\ell}{\ell!}$. Indeed, (17) can be further cast as

$$P_{\text{out}, k}^{\text{mmse}}(R) = F_{\eta_k}(2^R-1) - e^{-(2^R-1)/\delta} \sum_{\ell=0}^{n_r-n_t} \int_0^{2^R-1} \frac{e^{x/\delta} (2^R-1-x)^\ell}{\delta^\ell \ell!} f_{\eta_k}(x) dx, \quad (18)$$

whose first term can be evaluated from (7) by virtue of [14, 3.353.1], while the second requires the exploitation of [14, 3.196.1] to be expressed in closed form. As a consequence, we obtain

$$P_{\text{out}, k}^{\text{mmse}}(R) = 1 - e^{-(2^R-1)/\delta} \mathcal{K}_{n_t-1, n_r} \sum_{i=0}^{n_t-1} \frac{\mathcal{D}_i (n_r-i)!}{2^{R(n_r-i)}} \times \left\{ \frac{1}{2^R} + \sum_{\ell=0}^{n_r-n_t} \frac{(2^R-1)^{\ell+1}}{2^{R\ell} (\ell+1)! \delta^{\ell+1}} [\delta (n_r-i+1) \times {}_2F_1(\ell+1, \xi+1; \ell+2; 1-2^R) - {}_2F_1(\ell+1, \xi; \ell+2; 1-2^R)] \right\} \quad (19)$$

where $\xi = \ell - n_r + i + 1$, while \mathcal{K}_{n_t-1, n_r} and \mathcal{D}_i are defined in Proposition 4.1. Ongoing numerical investigation aims to identify the values of δ beyond which the high-SNR MMSE OP approximation [1, Formula (53)] and our exact result (19) closely match.

The corresponding expression for rank 1 and full rank Rice-faded interferers, can be obtained by plugging [4, Formula (51)] and [4, Formula (53)], respectively, in (17), while the derivation of an equivalent formula for the case of arbitrary Rician rank is subject of ongoing work. Unfortunately, no closed-form expression for $P_{\text{out}, k}^{\text{mmse}}(R)$ is available in the multiple-scattering case [13], due to the absence of any expression for $F_{\gamma_k^{\text{zf}}}$.

VI. NUMERICAL RESULTS

We now validate our analytical expressions for the pdf of η_k , against numerical (i.e., Monte Carlo) simulations.

Figure 1 shows analytical and numerical results, represented by markers and solid line, respectively. Results have been obtained for $n_t = 3$, $n_r = 5$ and different (finite) values of normalized SNR δ . The dashed line corresponds instead to the asymptotic expression of the pdf as $\text{SNR} \rightarrow \infty$, which is given in [1]. The plot highlights the excellent match between analysis and simulation, thus validating our derivations. Furthermore, we observe that the asymptotic expression provides a very good approximation of the pdf for values of normalized SNR greater than 10 dB, but it is far from providing an accurate representation for lower values of SNR.

In Figure 2 we show the pdf of the SNR gap η_k for the multiscattering channel and for a different numbers of scatterers and scattering stages. Again, the match between analysis and simulation is very tight. We observe that, as the number of scattering stages increases, the probability mass of f_{η_k} moves toward smaller values. Thus, the ZF performance approaches that of the MMSE receiver.

VII. CONCLUSION

The output SNR gap between linear MIMO MMSE and ZF has been investigated for arbitrary finite values of transmit power and finite number of antennas. Its pdf is provided in closed form, under the assumption of Rayleigh-faded intended stream and independent stream decoding. Interfering streams are, in turn, assumed to undergo Rayleigh, or Rician (with arbitrary rank and eigenvalues multiplicity) fading. The case of multiple Rayleigh scattering on the interfering streams is considered, too, thus providing a comprehensive analysis of fading scenarios, arising in current and foreseen wireless cellular settings. The extension beyond Rayleigh fading for the desired signal stream is currently under study.

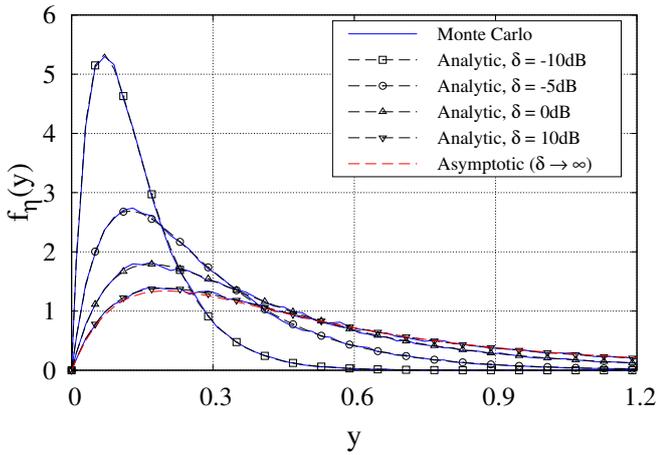


Fig. 1. Pdf of the SNR gap η_k for Rayleigh channel: comparison between analytical and numerical results, for $n_t = 3$, $n_r = 5$ and different values of normalized SNR δ . Results are also compared against the asymptotic expression [1].

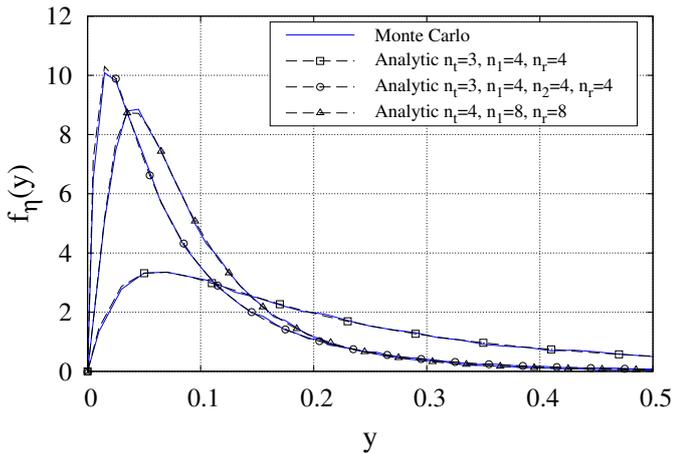


Fig. 2. Pdf of the SNR gap η_k for multiscattering channel: comparison between analytical and numerical results, for $n_t = 3$, $n_1 = 4$, $n_r = 4$; $n_t = 3$, $n_1 = 4$, $n_2 = 4$, $n_r = 4$; and $n_t = 4$, $n_1 = 8$, $n_r = 8$.

REFERENCES

- [1] Y. Jian, M. K. Varanasi, J. Li, "Performance Analysis of ZF and MMSE Equalizers for MIMO Systems: An In-Depth Study of the High SNR Regime," *IEEE Trans. on Inf. Th.*, Vol. 57, No. 4, pp. 6788–6805, 2011.
- [2] P. Li, D. Paul, R. Narasimhan J. Cioffi "On the Distribution of SINR for the MMSE MIMO Receiver and Performance Analysis," *IEEE Trans. on Inf. Th.*, Vol. 52, No. 1, pp. 271–286, 2006.
- [3] D. A. Gore, R. W. Heath, A. J. Paulraj, "Transmit Selection in Spatial Multiplexing Systems," *IEEE Comm. Lett.*, Vol. 6, No. 11, pp. 491–493, 2002.
- [4] C. Siriteanu, S. D. Blostein, A. Takemura, H. Shin, S. Yousefi, S. Kuriki, "Exact MIMO Zero-forcing Detection Analysis for Transmit-correlated Rician Fading," *IEEE Trans. on Wir. Comm.*, Vol. 13, No. 3, pp. 1514–1527, 2014.
- [5] C. Siriteanu, A. Takemura, S. Kuriki, D. St. P. Richards, H. Shin, "Schur Complement Based Analysis of MIMO Zero-forcing for Rician Fading," *IEEE Trans. on Wir. Comm.*, Vol. 14, No. 4, pp. 1757–1771, 2015.
- [6] H. Gao, P. J. Smith, M. V. Clark, "Theoretical Reliability of MMSE Linear Diversity Combining in Rayleigh-fading Additive Interference Channels," *IEEE Trans. on Comm.*, Vol. 46, No. 5, pp. 666–672, 2003.
- [7] M. R. McKay, A. Zanella, I. B. Collings, M. Chiani, "Error Probability and SINR Analysis of Optimum Combining in Rician Fading," *IEEE Trans. on Comm.*, Vol. 57, No. 3, pp. 676–687, 2009.
- [8] S. Jin, M. R. McKay, X. Gao, I. B. Collings, "MIMO Multichannel Beamforming: SER and Outage Using New Eigenvalue Distributions of Complex Noncentral Wishart Matrices," *IEEE Trans. on Comm.*, Vol. 56, No. 3, pp. 424–434, 2008.
- [9] R. Muller, "On the asymptotic eigenvalue distribution of concatenated vector-valued fading channels," *IEEE Trans. on Inf. Th.*, Vol. 48, No. 7, pp. 2086–2091, 2002.
- [10] G. Akemann, J. Ipsen, M. Kieburg, "Products of Rectangular Random Matrices: Singular Values and Progressive Scattering," *APS Phys. Rev. E*, Vol. 88, No. 3, 2013.
- [11] H. Shin, M. Z. Win, "MIMO diversity in the presence of double scattering," *IEEE Trans. on Inf. Th.* Vol. 54, No. 7, pp. 2976–2996, 2008.
- [12] C. Zhong, T. Ratnarajah, Z. Zhang, K.-K. Wong, M. Sellathurai, "Performance of Rayleigh Product MIMO Channels with Linear Receivers," *IEEE Trans. on Wir. Comm.*, Vol. 13, No. 4, pp. 2270–2281, 2014.
- [13] G. Alfano, C.-F. Chiasserini, A. Nordin, "Achievable Sum Rate of Linear MIMO Receivers with Multiple Rayleigh Scattering," *Proc. of WSA*, Munich, Mar. 9–11, 2016.
- [14] I. S. Gradshteyn, I. M. Ryzhik, *Table of Integrals, Series, and Products*, Academic Press, New York, 1980.
- [15] A. T. James, "Distribution of Matrix Variates and Latent Roots Derived from Normal Samples," *Ann. Math. Stat.*, Vol. 35, No. 2, pp. 474–501, 1964.
- [16] S. Verdú, *Multiuser Detection*, Cambridge University Press, 2011.
- [17] P. J. Smith, L. M. Garth, S. Loyka, "Exact Capacity Distributions for MIMO Systems with Small Numbers of Antennas," *IEEE Comm. Lett.*, Vol. 7, No. 10, pp. 481–483, 2003.
- [18] G. Alfano, C.-F. Chiasserini, A. Nordin, S. Zhou, "Closed-Form Output Statistics of MIMO Block-Fading Channels," *IEEE Trans. on Inf. Th.*, Vol. 60, No. 12, pp. 7782–7797, 2014.
- [19] M. Chiani, M. Z. Win, A. Zanella, "On the Capacity of Spatially Correlated MIMO Rayleigh-fading Channels," *IEEE Trans. on Inf. Th.*, Vol. 49, No. 10, pp. 2363–2371, 2003.
- [20] G. Alfano, C.-F. Chiasserini, A. Nordin, "Comparative Analysis of Linear MIMO Receivers under Various Fading Statistics," *Technical Report*, https://www.dropbox.com/s/3heax3yp71h46p0/Technical_Report.pdf?dl=0, Jan. 2016.
- [21] M. Matthaiou, P. de Kerret, G. K. Karagiannidis, J. A. Nossek, "Mutual Information Statistics and Beamforming Performance Analysis of Optimized LoS MIMO Systems," *IEEE Trans. on Wir. Comm.*, Vol. 58 No. 11, pp. 3316–3329, 2010.
- [22] A. Kuijlaars, D. Stivigny, "Singular Values of Products of Random Matrices and Polynomial Ensembles," *Random Matrices: Theory and Application*, Vol. 3, No. 3, pp. 1–22, 2014.