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Original

Multiple Bloch surface wave excitation with gratings / Lewis, Asilevi; Descrovi, Emiliano; Pesonen, Henri; Roussey, Matthieu; Turunen, Jari. - In: JOURNAL OF THE EUROPEAN OPTICAL SOCIETY. RAPID PUBLICATIONS. - ISSN 1990-2573. - 20:1(2024). [10.1051/jeos/2024007]

Availability:

This version is available at: 11583/2986256 since: 2024-02-23T09:02:12Z

Publisher:

EDP Sciences

Published

DOI:10.1051/jeos/2024007

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Multiple Bloch surface wave excitation with gratings

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Abstract. We study the coupling of a finite number of Bloch surface waves (BSWs) propagating in different directions at the surface of a dielectric multilayer. These surface waves arise from a set of diffraction orders associated to a grating on the bottom surface of the substrate that is illuminated by a normally incident beam. Simultaneous excitation of multiple BSWs is possible with a set of diffraction orders having the same radial spatial frequency. Using rigorous electromagnetic theory, we design gratings for simultaneous excitation of two, four and six BSWs propagating in directions separated by π , $\pi/2$ and $\pi/3$ azimuthal intervals, respectively.

Keywords: Evanescent waves, Surface Electromagnetic Waves, Bloch Surface Waves, Multiple Bloch Surface Waves, MBSWs.

1 Introduction

Surface Electromagnetic Waves (SEW) represent an interesting option for controlling optical signals on miniaturized chips for integrated optics and sensing applications. Surface Plasmon Polaritons (SPP) are probably the most widely known SEWs, but they exhibit inherent issues related to the ohmic losses introduced by the metallic materials involved. As an alternative, SEWs sustained by dielectric multilayers (ML) have attracted a growing interest in the past decade. This kind of SEW [1] are also referred to as Bloch Surface Waves (BSWs) [2] to highlight the role of the underlying periodic multilayer structure required for their existence. BSWs offer several advantages as compared to SPPs, such as a wide spectral tunability and low losses thanks to the large choice of transparent dielectric materials available for multilayer manufacturing. In addition, BSWs can be either TE- or TM-polarized [3,4], depending on the multilayer design. Their excitation by pulsed fields has also been recently studied numerically [5].

Being surface waves, BSWs are evanescent in the medium above the multilayer surface. The coupling with free-space radiation in a BSW-based device is therefore critical as it must provide momentum matching beyond the light-line. In most of the applications proposed so far, BSW coupling is performed by means of bulky prisms, either in Kretschmann or Otto configuration [6]. However, more sophisticated approaches have been recently implemented, involving, for example, the use of individual scatterers [7–9] or miniature prisms [10] placed onto the multilayer surface. Another promising option is represented by integrating diffraction gratings within the BSW-supporting structure [11]. This has been done mainly in two different ways: with the grating being fabricated on top of the multilayer [12–14] or buried beneath the multilayer [15]. In the first case, the multilayer is substantially planar, with the ex-

ception of the top layer, where the grating unavoidably perturbs the dispersion of the BSW mode (dielectric loading/unloading effect). In the second case, the grating is fabricated on the substrate surface prior to the multilayer deposition, which occurs on the same side. The multilayer itself results to not be perfectly planar because it (partially) conforms to the underlying corrugation. In both configurations, the BSW dispersion is altered by the presence of the grating, which may lead to some difficulties regarding the precise control of optical functions of complex, possibly resonating, BSW-based architectures.

We propose an alternative approach on diffractive coupling for BSWs, with gratings fabricated on the bottom surface of a transparent substrate having the multilayer deposited on the top surface. In particular, we explore the possibility of using two-dimensional gratings to simultaneously couple BSWs propagating in more than two directions by exploiting the momentum distribution of several diffraction orders. Once the mode dispersion of the multilayer is known, our approach facilitates BSW coupling in a controllable way, as far as wavelengths/numbers and propagation directions are concerned. The directional coupling of BSWs has been already tackled in a few previous articles [16–18], although never considered for multiple directions at once. When the optical path through the substrate is also taken into account, our approach allows a predictable control onto the coupling locations of BSWs launched in different directions.

The present paper is composed as follows. We begin, in Sec. 2, by introducing the grating-based BSW excitation principle and the assumed geometrical configuration. The theoretical framework for grating design, for which we use a rigorous technique known as the Fourier Modal Method (FMM) [19], is described in Sec. 3. The design process is analogous with the synthesis of grating-based mul-

multiple free-space beam splitters [20], but here we need to account for the BSW excitation conditions and the polarization state of the input wave. In Sec. 4, we first consider BSW stack design, providing a ‘benchmark’ stack employed in the rest of the work, and then cover the design of linear gratings for simultaneous excitation of two counter-propagating BSWs. Such designs are extended in Sec. 5 to two-dimensional periodic gratings for excitation of either four or six BSWs propagating at 90° or 60° intervals along the stack, respectively. After a discussion presented in Sec. 6, conclusions are drawn in Sec. 7.

2 Excitation principle and geometry

Figure 1 illustrates the geometry for the simplest case of excitation of two counter-propagating BSWs. A flat fused silica substrate with refractive index $n_{\text{sub}} = 1.462$, such as a 0.5 mm or 3 mm-thick SiO_2 plate, is illuminated by a normally incident monochromatic beam (wavelength λ_0) from the medium underneath (air). A linear grating, with period d of the order of λ_0 , provided on the air-substrate surface, splits the beam into three transmitted orders propagating within the substrate: the zeroth order $m = 0$ and the first diffracted orders $m = \pm 1$. The orders $m = \pm 1$ propagate in directions θ_{+1} and θ_{-1} given by $\sin \theta_{\pm 1} = \pm \lambda_0 / n_{\text{sub}} d$ towards the multilayer stack on the top surface of the substrate. If $|\theta_{\pm 1}|$ matches the Kretschmann-incidence BSW excitation angle θ_{BSW} for the given wavelength and polarization state (TE or TM), two counter-propagating BSWs are generated simultaneously. The excitation is efficient as long as the angular spectrum of each diffracted order, which defines the beam divergence, falls essentially within the (stack-dependent) BSW momentum bandwidth. The polarization state of illumination affects the coupling significantly; we will consider only BSW excitation in TE polarization, which generally requires a smaller number of stack layers than TM-polarized BSW excitation.

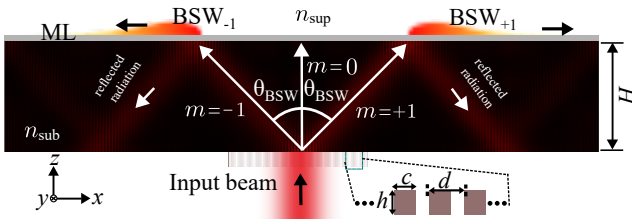


Figure 1: Principle of MBSW generation: the two-beam case. A binary linear surface-relief grating defined by period d , ridge width c , and ridge height h on the bottom of a substrate of thickness H splits the input beam into two diffracted orders $m = -1$ and $m = +1$, which excite BSWs on the top surface of the substrate by interaction with the multilayer stack (ML). We assumed $n_{\text{sub}} = 1.462$ and n_{sup} is air.

The parameters of the system are chosen such that the two BSWs shown in Fig. 1 are spatially separated under finite-beam illumination. This feature can be useful in BSW-based platforms such as interferometers [23] and integrated components [24]. First-order diffracted beams are partially reflected at the top interface, thus propagating back into the substrate. The reflected beams continue to propagate according to multiple-reflection paths inside the sub-

strate unless they are extracted by means of diffusers or gratings. At each reflection with the ML interface, coupling to BSW occurs. Stated differently, BSWs are launched at different locations on the ML surface each time the beam is incident on the bottom interface of the dielectric stack, thus leading to the appearance of BSW interference effects unless the substrate thickness H is sufficiently large to minimize spatial overlaps.

3 Theoretical framework

Let us consider a rectangularly periodic grating of period $d_x \times d_y$ in the cartesian xy coordinate system and assume a plane-wave illumination (at frequency ω) normally incident onto the substrate from air. In view of the grating equations, the wave vectors of the propagating diffraction orders (m, n) in the substrate are

$$\mathbf{k}_{mn} = k_{xm}\hat{x} + k_{yn}\hat{y} + k_{zmn}\hat{z} \quad (1)$$

where

$$k_{xm} = mK_x = 2\pi m/d_x, \quad (2a)$$

$$k_{yn} = nK_y = 2\pi n/d_y, \quad (2b)$$

$$k_{zmn} = \sqrt{k_0^2 n_{\text{sub}}^2 - k_{xm}^2 - k_{yn}^2}, \quad (2c)$$

and $k_0 = \omega/c_0 = 2\pi/\lambda_0$ is the wave number in vacuum. After defining the radial spatial frequency of the generic order (m, n) as

$$k_{pmn} = \sqrt{k_{xm}^2 + k_{yn}^2} = \sqrt{(mK_x)^2 + (nK_y)^2}, \quad (3)$$

the condition $k_{pmn} < k_0 n_{\text{sub}}$ identifies those diffraction orders propagating within the substrate, the others being evanescent. If we denote the refractive index of the superstrate by n_{sup} and assume $n_{\text{sup}} < n_{\text{sub}}$, order (m, n) is evanescent in the superstrate when $k_{pmn} > k_0 n_{\text{sup}}$. Considering BSW excitation, we are therefore interested in orders with radial spatial frequencies in the range $k_0 n_{\text{sup}} < k_{pmn} < k_0 n_{\text{sub}}$. We are primarily interested in the nearest neighbors of the zeroth transmitted order, while higher orders are made evanescent by appropriate choices of d_x and d_y . In the illustrative example presented in Fig. 2(a), orders $(m, n) = (-1, 0)$ and $(m, n) = (+1, 0)$ fall on the yellow line of radius $k_{\rho\text{BSW}}$, which defines the BSW excitation condition dictated by the ML design.

Following Ref. [21], we define the ‘exit plane’ of diffraction order (m, n) as the plane containing the wave vector \mathbf{k}_{mn} and the unit vector \hat{z} . Further, propagation angles θ_{mn} and ϕ_{mn} of the transmitted orders, are defined as

$$k_{xm} = k_0 n_{\text{sub}} \sin \theta_{mn} \cos \phi_{mn}, \quad (4a)$$

$$k_{yn} = k_0 n_{\text{sub}} \sin \theta_{mn} \sin \phi_{mn}, \quad (4b)$$

$$k_{zmn} = k_0 n_{\text{sub}} \cos \theta_{mn}. \quad (4c)$$

as illustrated in Fig. 2(b). Here ϕ_{mn} is the azimuthal angle in the range $[0, 2\pi)$, measured counter-clockwise from the k_x axis, and θ_{mn} in the range $[0, \pi/2)$ is the propagation angle measured from

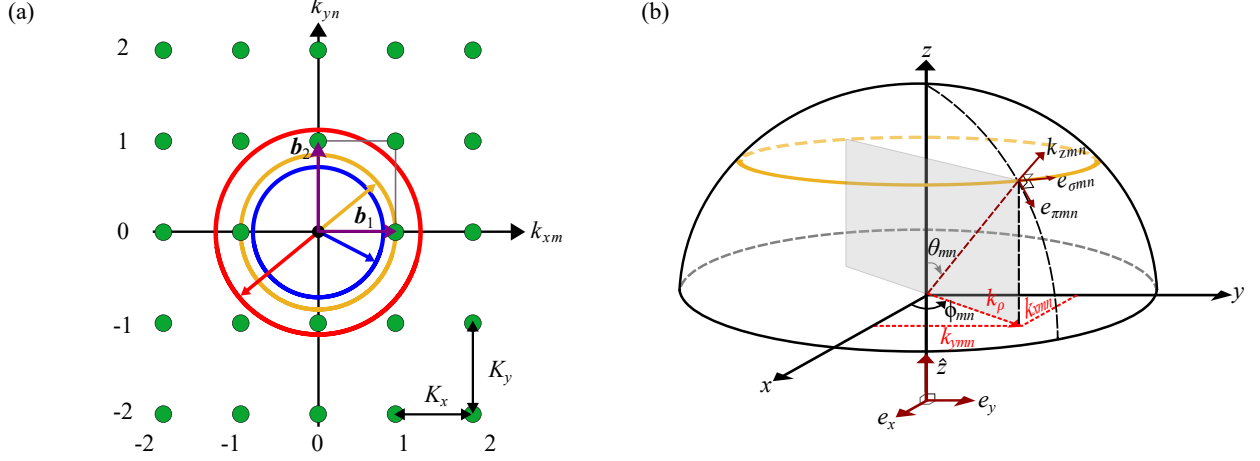


Figure 2: (a) Diffraction orders of a rectangular lattice in spatial-frequency representation at normal incidence. Diffraction orders are represented by dots at positions $k_{xm} = mK_x$, $k_{yn} = nK_y$. Blue and red circles represent the cut-off radial spatial frequencies $k_\rho = k_0 n_{\text{sub}}$ and $k_\rho = k_0 n_{\text{sup}}$, respectively, between which BSW excitation is possible. The yellow circle indicates the radial spatial frequency of BSW on a given ML. (b) Definition of the propagation angles (θ_{mn} , ϕ_{mn}) of a single transmitted diffracted order (m, n) in its exit plane (the grey rectangle) and the $\pi - \sigma$ basis of the diffracted electric field.

the k_z axis. It will prove convenient to use the so-called $\pi - \sigma$ basis (or local TM/TE basis) to define the polarization states of the transmitted orders. As described in Ref. [21], this basis allows us to treat incident fields with any polarization state, including partial polarization. Here, however, we are mainly interested in either fully polarized or unpolarized illumination.

If the incident plane wave is fully polarized, we can use any suitable rigorous grating analysis method (in our case FMM) to determine the transverse Cartesian components e_{xmn} and e_{ymn} of the polarization vector for any transmitted order, as discussed shortly below. The longitudinal component of e_{mn} is fixed by Maxwell's divergence equation, which gives $\mathbf{k}_{mn} \cdot \mathbf{e}_{mn} = 0$ and

$$e_{zmn} = -\frac{1}{k_{zmn}} (k_{xm} e_{xmn} + k_{yn} e_{ymn}). \quad (5)$$

In the $\pi - \sigma$ basis the polarization state of any order is described by a two-dimensional vector $\mathbf{e}_{\pi\sigma mn} = [e_{\pi mn}, e_{\sigma mn}]^T$, where the π and σ components are explicitly given by

$$e_{\pi mn} = e_{xmn} \cos \theta_{mn} \cos \phi_{mn} + e_{ymn} \cos \theta_{mn} \sin \phi_{mn} - e_{zmn} \sin \theta_{mn}, \quad (6a)$$

$$e_{\sigma mn} = -e_{xmn} \sin \phi_{mn} + e_{ymn} \cos \phi_{mn}. \quad (6b)$$

As shown in Fig. 2(b), the component $e_{\pi mn}$ lies in the exit plane, whereas $e_{\sigma mn}$ is perpendicular to it. Hence, they represent the TM and TE components of the electric field in the exit plane, respectively.

In diffraction by two-dimensionally periodic gratings, the polarization states of the transmitted (and reflected) diffracted orders generally depend on the state of input polarization. We represent the polarization vector of a (generally, elliptically polarized) unit-amplitude input plane wave as

$$\mathbf{e} = e_x \hat{\mathbf{x}} + e_y \hat{\mathbf{y}} = \hat{\mathbf{x}} \cos \alpha + \hat{\mathbf{y}} \sin \alpha \exp(i\delta), \quad (7)$$

normalized such that $\mathbf{e} = 1$. The effect of the grating on transmitted radiation can be analyzed by calculating (by FMM) the transmis-

sion coefficients

$$T_{xmn}^{(x)}, T_{ymn}^{(x)}, T_{xmn}^{(y)}, T_{ymn}^{(y)}, \quad (8)$$

for all diffraction orders, where the superscripts (x) and (y) refer to illumination by a purely x -polarized ($\mathbf{e} = \hat{\mathbf{x}}$) or y -polarized ($\mathbf{e} = \hat{\mathbf{y}}$) incident wave. The coefficients in Eq. (8) are precisely the complex vector amplitudes that appear in the Rayleigh plane-wave expansion of the field at the output plane of the grating; see, e.g., Eq. (5) in Ref. [20]. For an arbitrarily (fully) polarized incident wave the transverse electric-field components of the transmitted orders are [21]

$$e_{xmn} = T_{xmn}^{(x)} e_x + T_{xmn}^{(y)} e_y, \quad (9a)$$

$$e_{ymn} = T_{ymn}^{(x)} e_x + T_{ymn}^{(y)} e_y. \quad (9b)$$

The longitudinal components e_{zmn} are obtained from Eq. (5), and the $\pi - \sigma$ representation of each order is given by Eqs. (6). Since the input polarization state affects both the π and σ components, it can be used as a design degree of freedom in multiple-BSW excitation, in addition to the geometrical grating parameters.

It is customary to describe the state of polarization of a fully polarized field by a 2×2 polarization matrix $\mathbf{J} = \mathbf{e}^* \mathbf{e}^T$ (Ref. [22], sec. 6.3.2). Explicitly, for the incident field,

$$\mathbf{J} = \begin{bmatrix} J_{xx} & J_{xy} \\ J_{yx} & J_{yy} \end{bmatrix} = \begin{bmatrix} |e_x|^2 & e_x^* e_y \\ e_y^* e_x & |e_y|^2 \end{bmatrix}, \quad (10)$$

where the asterisk denotes complex conjugation. Correspondingly, the polarization state of any transmitted order in the $\pi - \sigma$ basis is described by $\mathbf{J}_{\pi\sigma mn} = \mathbf{e}_{\pi mn}^* \mathbf{e}_{\sigma mn}^T$ [21], explicitly

$$\mathbf{J}_{\pi\sigma mn} = \begin{bmatrix} J_{\pi\pi mn} & J_{\pi\sigma mn} \\ J_{\sigma\pi mn} & J_{\sigma\sigma mn} \end{bmatrix} = \begin{bmatrix} |e_{\pi mn}|^2 & e_{\pi mn}^* e_{\sigma mn} \\ e_{\sigma mn}^* e_{\pi mn} & |e_{\sigma mn}|^2 \end{bmatrix}. \quad (11)$$

The polarization states of the diffracted orders can also be characterized by the Stokes parameters [21]

$$S_{0mn} = J_{\pi\pi mn} + J_{\sigma\sigma mn}, \quad (12a)$$

$$S_{1mn} = J_{\pi\pi mn} - J_{\sigma\sigma mn}, \quad (12b)$$

$$S_{2mn} = 2\Re(J_{\pi\sigma mn}), \quad (12c)$$

$$S_{3mn} = 2\Im(J_{\pi\sigma mn}), \quad (12d)$$

where \Re and \Im denote the real and imaginary parts. The normalized forms of the Stokes parameters are defined as $s_{jmn} = S_{jmn}/S_{0mn}$ ($j = 1, 2, 3$), and the degree of polarization associated with order (m, n) is given by

$$P_{mn} = \sqrt{s_{1mn}^2 + s_{2mn}^2 + s_{3mn}^2}. \quad (13)$$

For a fully polarized incident wave, $P_{mn} = 1$ for all orders, even though the values of the individual Stokes parameters generally depend on order indices.

In addition to fully polarized illumination, we consider the opposite extreme case of unpolarized illumination. The matrix \mathbf{J} for partially polarized light is defined as $\mathbf{J} = \langle \mathbf{e}^* \mathbf{e}^T \rangle$, where the brackets denote ensemble averaging over all polarization realizations. For unpolarized illumination it has a diagonal form (Ref. [22], sec. 6.3.3)

$$\mathbf{J} = \frac{1}{2} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \quad (14)$$

and the degree of input polarization is $P = 0$. The polarization matrix associated with order (m, n) can be represented as an average

$$\mathbf{J}_{\pi\sigma mn} = \frac{1}{2} [\mathbf{J}_{\pi\sigma mn}^{(x)} + \mathbf{J}_{\pi\sigma mn}^{(y)}], \quad (15)$$

which remains diagonal because e_x and e_y are uncorrelated. However, since the grating treats these components differently, in general $J_{\pi\pi mn} \neq J_{\sigma\sigma mn}$, implying that the individual orders become partially polarized with $P_{mn} > 0$.

In standard beam splitting problems in resonance-domain diffractive optics [20] one is interested in the distribution of the diffraction efficiencies of the propagating orders. Since we have normalized the intensity of the incident field such that $S_0 = 1$, the diffraction efficiencies are defined as [21]

$$\eta_{mn} = n_{\text{sub}} \cos \theta_{mn} S_{0mn}. \quad (16)$$

In BSW excitation problem, the design goal is to maximize and equalize the coupling of the incident field to a set of BSW modes with the angle θ_{mn} equal to θ_{BSW} . If θ_{BSW} is the excitation angle for TE polarization, the component $e_{\sigma mn}$ excites a BSW while $e_{\pi mn}$ is non-resonant, and vice versa.

We choose the geometry such that several diffraction orders have the same radial spatial frequency $k_{\rho mn} = k_0 n_{\text{sub}} \sin \theta_{\text{BSW}}$, and therefore lie on the yellow circle depicted in Fig. 2(a). The relative amplitudes of the excited BSWs are determined by the σ -polarized

components $J_{\sigma\sigma mn}$ in TE polarization and $J_{\pi\pi mn}$ in TM polarization. The fraction

$$\kappa_{mn} = J_{\sigma\sigma mn} / J_{\pi\pi mn} \quad (17)$$

provides the ratio of the coupled and uncoupled parts of the incident wave in BSW excitation.

4 Plane-wave design with linear gratings

As evident from the preceding discussion, the $\pi - \sigma$ representation returns the BSW excitation problem to the basic TE or TM polarized problem. In addition, since we assume a substrate thickness $H \gg \lambda_0$, the evanescent parts of the diffracted fields above the grating and the BSW field below the stack are spatially well separated. Hence, we may treat the BSW stack design and the grating design as two separate problems. In order to obtain an illustrative stack design useful for our purposes, we fix $\lambda_0 = 514$ nm and provide a stack geometry sustaining BSWs at angles between the blue and red lines in Fig. 2(a). The resulting stack can then be used to design gratings for excitation of BSWs that lie on the yellow circle in Fig. 2(a).

4.1 Multilayer stack design

Figure 3(a) shows the assumed stack structure, which consists of N high/low (H/L) refractive index bilayers and a terminating top (T) layer with refractive indices n_{H} , n_{L} , n_{T} and thicknesses h_{H} , h_{L} , h_{T} , respectively. To reduce the number of variable parameters, we consider TE polarization, fix the number of bilayers to $N = 6$, use refractive indices $n_{\text{H}} = 2.520$ (TiO₂), $n_{\text{L}} = 1.476$ (SiO₂), $n_{\text{T}} = n_{\text{H}} = 2.520$. The thicknesses h_{H} , h_{L} , h_{T} are used to design the stack such that the BSW resonance occurs at an angle θ_{mn} in the exit plane.

Figure 3(b) shows the design results. The horizontal axis is $k_x/k_0 = n_{\text{sub}} \sin \theta_{mn} = n_{\text{eff}}$, where n_{eff} can be interpreted as the effective index of the stack. The plotting range starts from the critical angle of BSW generation and extends to $k_x/k_0 n_{\text{sub}}$, i.e., it spans the region between the blue and red circles in Fig. 2(a). As n_{eff} increases, the BSW becomes increasingly buried within the multilayer and acts less like a surface mode. At the same time, all layer thicknesses show a monotonically increasing trend.

4.2 Two-way splitting

As illustrated in Fig. 1, the coupling of two counter-propagating BSWs is possible with linear gratings ($d_x = d$, $d_y = \infty$). The exit plane of both orders, $(m, n) = (-1, 0)$ and $(+1, 0)$, is the xz plane and the $\pi - \sigma$ representation reduces to the standard TM/TE

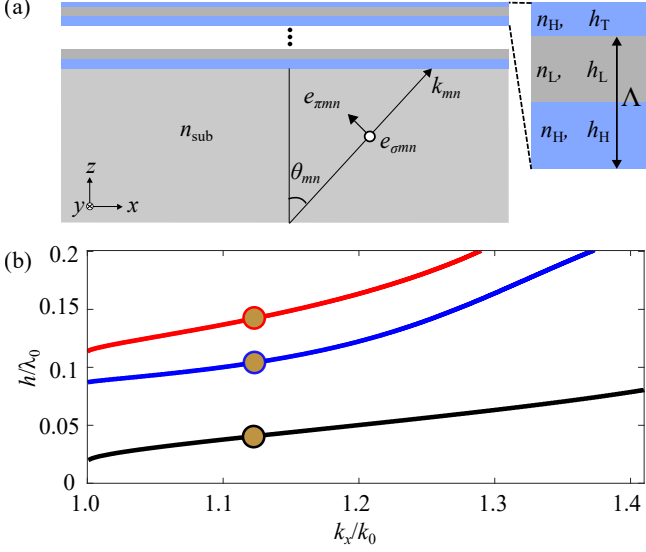


Figure 3: (a) Definition of the multilayer structure and notation. (b) Stack parameters as a function of the ratio $k_x/k_0 = n_{\text{eff}}$ for TE-mode BSW excitation with $N = 6$ bilayers: h_H/λ_0 (blue), h_L/λ_0 (red), and h_T/λ_0 (black). The dots mark the position $k_x/k_0 = 1.1209$ for BSW excitation at 50° angle of incidence.

decomposition. Since, by symmetry, $\eta_{-1,0} = \eta_{+1,0}$ for binary profiles defined in the inset of Fig. 1, we need to maximize $\eta_{+1,0}$. This also leads to the optimum value of $J_{\sigma\sigma mn}$, while $J_{\pi\pi mn} = 0$. Now we only need to find the values of the fill factor $f = c/\lambda_0$ and grating height h/λ_0 that maximize $\eta_{+1,0}$ ($= \eta_{-1,0}$) to also maximize $J_{\sigma\sigma}$.

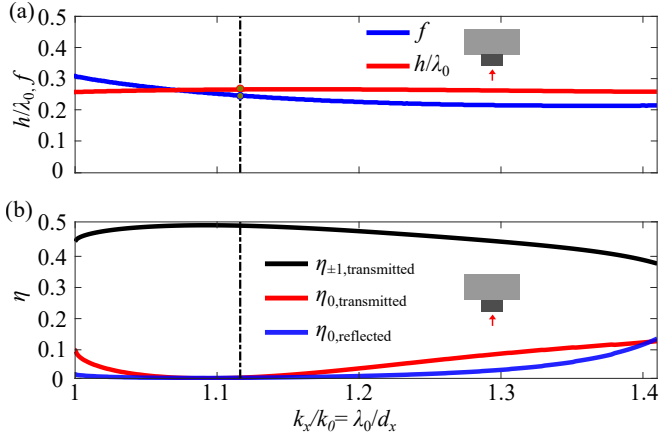


Figure 4: Design of two-way beam splitters. (a) Optimum values of the fill factor f (red) and the relief depth h/λ_0 (blue) as a function of the exit angle of the first diffracted order in TE polarization. (b) The corresponding first-order diffraction efficiency $\eta_{+1,0} = \eta_{-1,0}$ (black), the efficiency $\eta_{0,0}$ of the zeroth transmitted order (red), and that of the zeroth reflected order (blue), which is $1 - 2\eta_{+1,0} - \eta_{0,0}$ due to energy conservation. The inset shows the grating structure and direction of illumination.

The grating-design results are summarized in Fig. 4(a). The optimum fill factor remains fairly constant over the entire angular range considered here, whereas the optimum grating height decreases with increasing angle. The efficiencies of all propagating orders are plotted in Fig. 4(b). At around $k_x/k_0 = 1.1209$ (corresponding to an excitation angle 50°) we get $\eta_{\pm 1,0} \approx 0.4973$. Some

light is ‘lost’ in zeroth reflected and transmitted orders when we move close to the cut-off at $k_x/k_0 = 1$ or towards larger values of k_x/k_0 , but the designs remain acceptable over a relatively wide range of excitation angles.

The results in Figs. 3(b) and 4 allow us to design two-way beam splitters for any BSW resonance angle of interest. The stack design for the desired angle is obtained from Fig. 3(b) and the corresponding grating design from Fig. 4(a). The performance of the design can be evaluated from Fig. 4(b). To limit the number of variables further, we set $\theta_{\text{BSW}} = 50^\circ$, corresponding to $k_x/k_0 = 1.1209$. The stack design is marked by the dots in Fig. 3(b), the optimum parameters for TE excitation with $N = 6$ bilayers being $h_H = 60$ nm, $h_L = 85$ nm, and $h_T = 20$ nm. Correspondingly, the vertical lines in Fig. 4(a) give a grating design $f = 0.2536$, $h/\lambda_0 = 0.2752$, with $\eta_{\pm 1,0} = 0.4973$.

Considering the optimized case represented by the dots on Fig. 3 and the black dashed line on Fig. 4, simulation of the reflected and transmitted coefficients has been performed for the full structure. It implies a grating of period $d \simeq 459$ nm and fill factor is $f = 0.2536$ and $h/\lambda_0 = 0.2752$ on the lower side of a $10\text{-}\mu\text{m}$ -thick fused silica wafer on top of which the multilayer is deposited. The multilayer design leads to a Bloch surface wave excited when the first diffracted order emerge from the grating at an angle of 50° ($k_x/k_0 = 1.1209$) at a wavelength of 514 nm. This is observed in Fig. 5(a), where a strong dip in reflection arises at this value of k_x/k_0 . In Fig. 5(b) the response in wavelength is presented. One can observe a relatively strong peak in transmission slightly shifted with regards to the reflection dip.

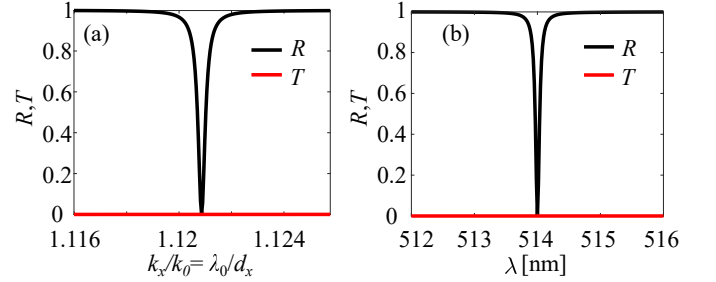


Figure 5: Response of the full structure (grating, substrate, multilayer and superstrate). (a) and (b) Reflected (black curves) and transmitted (red curves) first diffracted orders as a function of the normalized wavevector (a) and wavelength (b).

5 Plane-wave design with biperiodic gratings

We proceed to design of two-dimensionally periodic gratings that allow simultaneous excitation of more than two BSWs. Two lattice geometries are considered: square lattices for four-way excitation and hexagonal lattices for six-way excitation.

5.1 Four-way splitting

Let us first consider biperiodic gratings with primitive direct-lattice vectors $\mathbf{a}_1 = d\hat{x}$, $\mathbf{a}_2 = d\hat{y}$. The (Wigner–Seitz) primitive cell is square-shaped, covering the area $-d/2 < x < d/2$, $-d/2 < y < d/2$. The spatial frequencies of the diffraction orders are then $k_{xm} = mK$, $k_{yn} = nK$, the coordination number is 4, and the nearest neighbors of the zeroth order $(0, 0)$, namely $(m, n) = (+1, 0)$, $(0, +1)$, $(-1, 0)$, $(0, -1)$, propagate in directions $\phi_{+1,0} = 0$, $\phi_{0,+1} = \pi/2$, $\phi_{-1,0} = \pi$, $\phi_{0,-1} = 3\pi/2$, respectively. By an appropriate choice of d , all of these four orders can be placed simultaneously on the yellow ring in Fig. 2(a), thus enabling four-way BSW excitation.

In the design, we found it sufficient to consider binary (z -invariant) relative-permittivity profiles of the particular form

$$\epsilon_r(x, y, z) = \begin{cases} n_1^2 & \text{when } x^2 + y^2 < r^2 \\ n_2^2 & \text{otherwise} \end{cases} \quad (18)$$

in $0 < z < h$ within the primitive cell. The circular feature defined by the radius r can be either a pillar ($n_1 = n_{\text{sub}}$, $n_2 = 1$ or a hole ($n_1 = 1$, $n_2 = n_{\text{sub}}$) etched in the substrate. This type of pillar/hole structures can be patterned at a nanometer-scale addressing resolution using electron beam lithography system available to us, and require only a single etching step.

The radius r and the relief depth h can be used as the structural design parameters. Some symmetry rules exist, which are helpful in the design. Since the unit cell and the structure are centered at the origin, the transmission coefficients in Eq. (8) satisfy the inversion symmetry rules

$$T_{x,-m,-n} = T_{xmn}, \quad T_{y,-m,-n} = T_{ymn}, \quad (19)$$

for both (x) and (y) input polarizations. These rules hold regardless of the input polarization state, which however has an effect on the actual values of T_{xmn} and T_{ymn} . They reduce the number of orders that we need to (or can) control from four to two: we see from Eq. (19) that $\eta_{-m,-n} = \eta_{mn}$. Similar symmetry rules hold also for $J_{\sigma\sigma mn}$ and $J_{\pi\pi mn}$.

We begin the design of four-way couplers by optimizing the structural parameters r and h to minimize the sum of the efficiencies of the reflected and transmitted zeroth orders. This maximizes the combined efficiency of the four nearest-neighbor diffraction orders, and leaves the polarization state of the incident field free for design. Choosing $\theta_{\text{BSW}} = 50^\circ$ ($d \approx 0.892\lambda_0$), for either 45° or circularly polarized illumination and considering pillars, we get a design $r \approx 0.201\lambda_0$, $h \approx 0.53176\lambda_0$, which gives reflected and transmitted zero-order efficiencies of $\sim 3.5\%$ and $\sim 5.2\%$, respectively, leaving the rest of the incident energy to be distributed among the nearest-neighbor orders.

Our remaining target is to equalize (and maximize) the coupling strengths $J_{\sigma\sigma mn}$ of the four signal orders by designing the input polarization state defined in Eq. (7). The symmetry in the 4-way

splitting implies that there is no structurally induced polarization conversion: for (x) -polarized input we get $J_{\sigma\sigma mn} = 0$ for orders $(m, n) = (\pm 1, 0)$, while (y) -polarized input gives $J_{\sigma\sigma mn} = 0$ for orders $(m, n) = (0, \pm 1)$. Considering linearly polarized light, the values of $J_{\sigma\sigma mn}$ (and $J_{\pi\pi mn}$) vary rapidly with the angle α . Choosing $\alpha \approx \pi/4$ gives values $J_{\sigma\sigma mn} \approx 0.063$ and $\kappa_{mn} \approx 0.359$ for all four orders. The same result is obtained also for circularly polarized illumination with $\alpha \approx \pi/4$, $\delta = \pm\pi/2$. Both the optimized diffracted efficiencies and the maximized coupling strengths occur at the same illumination polarization.

Considering unpolarized illumination, the matrix $\mathbf{J}_{\pi\sigma mn}$ becomes diagonal and the degree of polarization takes the form

$$P_{mn} = |s_{1mn}| = \frac{|J_{\pi\pi mn} - J_{\sigma\sigma mn}|}{J_{\pi\pi mn} + J_{\sigma\sigma mn}}. \quad (20)$$

With the present numerical values we obtain $P_{mn} \approx 0.473$ for all nearest-neighbor orders. Even though the excitation wave is partially polarized, we obtain the same values of $J_{\sigma\sigma mn}$ as above; both of the two mutually uncorrelated components of the incident field contribute to TE-mode BSW excitation.

5.2 Six-way splitting

Let us consider a grating with hexagonal symmetry, which allows simultaneous excitation of six BSWs. The primitive vectors are now $\mathbf{a}_1 = d\hat{x}$ and $\mathbf{a}_2 = (d/2)\hat{x} + (\sqrt{3}d/2)\hat{y}$, and the Wigner–Seitz primitive cell is a hexagon as shown in Fig. 6(a). It will, however, be convenient for our purposes to define a rectangular direct-lattice cell as in Ref. [20], which covers the spatial region $-d/2 < x < d/2$, $-\sqrt{3}d/2 < y < \sqrt{3}d/2$ in Fig. 6(a). This alternative lattice representation simplifies the visualization of the geometry. It also allows the use of FMM in a cartesian instead of a non-orthogonal basis, as in Ref. [20], though in the present work we actually used the latter basis.

The spatial-frequency structure defined by the reciprocal-lattice primitive vectors $\mathbf{b}_1 = K\hat{x} - (K/\sqrt{3})\hat{y}$, $\mathbf{b}_2 = (2K/\sqrt{3})\hat{y}$ with $K = 2\pi/d$ is illustrated in Fig. 6(b), where the solid green circles show the locations of the allowed orders in the cartesian (k_x, k_y) system. The empty circles represent the orders of the rectangular spatial lattice, which are forbidden by the hexagonal symmetry. The yellow circle connects the six nearest neighbors of the zeroth order that satisfy the condition for BSW excitation simultaneously: orders $(m, n) = (+1, +1)$, $(0, +2)$, $(-1, +1)$, $(-1, -1)$, $(0, -2)$, $(+1, -1)$ of the rectangular lattice, with exit planes at angles $\pi/6 + q\pi/3$, $q = 0, \dots, 5$. The excited BSWs propagate along the surface of the stack in these directions.

In hexagonal lattice geometry, the symmetry rules in Eq. (19) ensure $J_{\sigma\sigma,-1,-1} = J_{\sigma\sigma,1,1}$, $J_{\sigma\sigma,0,-2} = J_{\sigma\sigma,0,2}$, $J_{\sigma\sigma,-1,1} = J_{\sigma\sigma,1,-1}$. These symmetries leave us three pairs of orders to control, and we expect to need additional structural freedom compared to the 4-wave case. Let us nevertheless see what designs are pos-

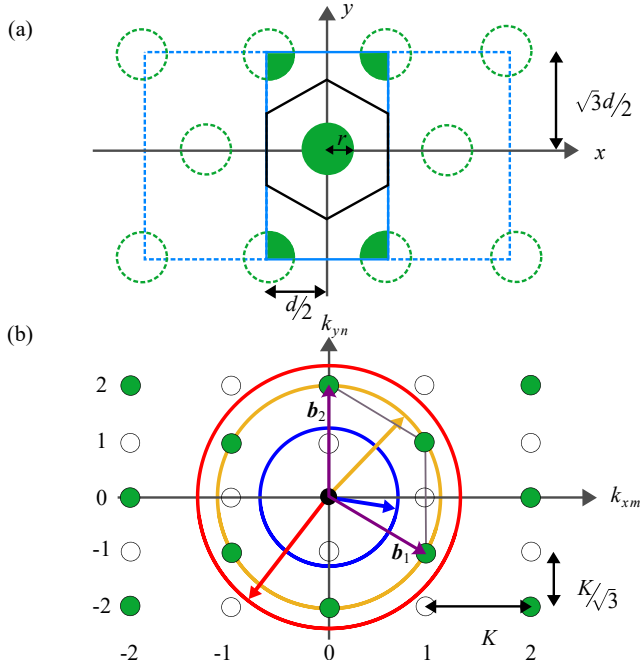


Figure 6: (a) The spatial structure of a hexagonal grating. The hexagon shows the spatial Wigner-Seitz primitive cell, the green features illustrate the grating structure, and the blue rectangle shows the non-primitive cartesian cell. (b) The spatial-frequency grid. The filled and empty dots represent the allowed and forbidden orders of the hexagonal lattice. The blue, yellow, and red circles have the same meaning as in Fig. 2(a).

sible with circular pillars by following the same strategy as above. An important difference is that in the hexagonal geometry we do have structurally induced polarization conversion.

By optimizing r and h for pillars, we get $h \approx 0.422\lambda_0$ and $r \approx 0.254\lambda_0$, which leaves a combined efficiency of ~ 0.884 available for the 6 orders of interest. The distribution of $J_{\sigma\sigma mn}$ again depends on input polarization. We found that it is not possible to equalize the coupling exactly for all six orders, but using circularly polarized light with $\alpha = \pi/4$ and $\delta = 0.486\pi$ we have $J_{\sigma\sigma,1,1} = 0.097$, $J_{\sigma\sigma,0,2} = 0.102$, and $J_{\sigma\sigma,1,-1} = 0.113$, respectively. Similarly, for κ_{mn} , we have $\kappa_{1,1} = 1.691$, $\kappa_{0,2} = 1.903$, and $\kappa_{1,-1} = 2.310$. Though it is not of concern for the present purposes, it is worth noting that the diffraction efficiencies are: $\eta_{1,1} \approx 0.152$, $\eta_{0,2} \approx 0.145$, and $\eta_{1,-1} \approx 0.144$. As with the four-wave case, the design with circular pillars works also for circularly polarized or unpolarized illumination but the exact values of $J_{\sigma\sigma mn}$ depend on polarization, but are within the same range as above. For unpolarized illumination, the degrees of polarization of the individual orders are nearly equal, $P_{mn} \approx 0.3226$.

In Fig. 7, we show the field amplitude distribution associated with the six-way coupling geometry in the xz -plane, i.e., crossing the multilayer, when illuminated with a 45° polarized light wave. The field is evaluated over 3-unit cells, i.e., 3 grating periods, in the x -direction. It shows, as expected, a strong field on the upper medium, which such structure ideal for sensing applications, espe-

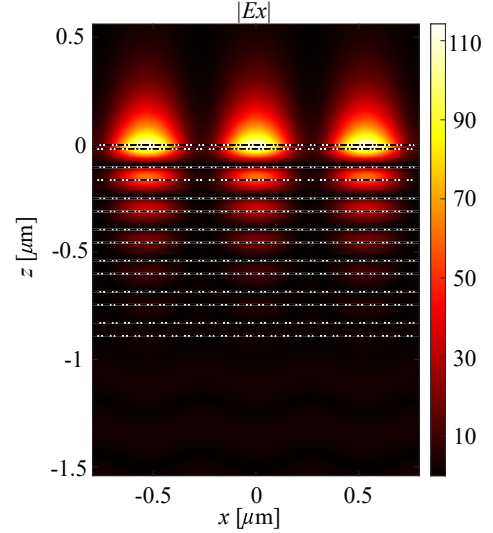


Figure 7: Field amplitude distribution across the multilayer (xz -plane). The illumination polarization was set to 45° . The dashed lines superimposed on the field represent the multilayer interfaces.

cially when providing multiple sensing areas thanks to the splitting of the BSW excitation.

6 Discussion

Throughout the paper we have considered normally incident illumination. The use of non-normal incidence could potentially allow us to consider other combinations of diffracted orders being simultaneously resonant. Changing the angle of incidence moves the grid of diffracted orders transversely in Fig. 2(a) with respect to the circles centered at the origin. For instance, if the propagation direction of the incident field is chosen as $(k_{xi}, k_{yi}) = (0, k_{yi})$, increasing k_{yi} moves the grid downwards in k_y direction, giving the orders at positions $k_{xm} = mK_x$, $k_{yn} = k_{yi} + nK_y$. Hence the three orders $(m, n) = (0, 0)$ and $(m, n) = (\pm 1, 0)$ would have a common radial spatial frequency if $k_{yi} = -(K_x^2 + K_y^2)/2K_y$, being therefore available for 3-way BSW excitation. To avoid order $(0, 2)$ from occupying the same ring as the zeroth order, we would need to choose $K_x \neq K_y$. However, placing the (yellow) BSW resonance ring outside the blue ring in Fig. 2(a) requires $k_{p00} > k_0 n_{\text{sub}}$, which is not possible with incidence from air. Hence a Kretschmann excitation geometry would be needed, thus sacrificing the compact footprint of the setup.

As an alternative to the geometry considered in this paper, we could consider having the splitter grating and the BSW stack as an integrated structure. This would still allow a compact platform at normal incidence, but the grating design and BSW stack design would not be independent anymore. As a first drawback, the splitter grating would most likely have to be rather thick ($\sim \lambda_0$) to suppress the zeroth transmitted order, preventing the possibility of etching it in the top ML layer. As a consequence, a strong degrading effect in the excited BSWs would be expected. Alternatively, one could use a highly index-modulated splitter grating with a flat top surface immediate below the stack. This would partially alleviate the

dependency in the ML and the grating design, but presumably the BSWs would be less affected.

7 Conclusions

In summary, we considered grating design for 2-way, 4-way, and 6-way BSW coupling at normally incident but arbitrarily polarized illumination of gratings with linear, square, and hexagonal symmetries. The plane-wave designs feature ideal TE-mode BSW coupling in the two-wave case. In the other cases the non-resonant parts of the excitation orders cannot be eliminated simultaneously, and they actually dominate the resonant (σ polarized) parts by a factor of ~ 2.8 in the 4-wave case. The 6-wave case reveals the opposite observation with the resonant part dominating by a factor of ~ 1.968 making them ideal candidate for TE-mode BSW excitation. Nevertheless, in the 4-wave case all four coupling ratios can be made equal, and in the 6-wave case practically equal, for several input polarization states of practical significance.

Funding

The work was partially funded by the Academy of Finland through project 333938 and the Flagship Program PREIN (346518). E. Descrovi acknowledges the funding received by Italian “Ministero dell’Università e della Ricerca” under the “Dipartimento di Eccellenza 2018-2022” program.

Disclosures

The authors declare no conflicts of interest.

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